

Amsterdam Machines

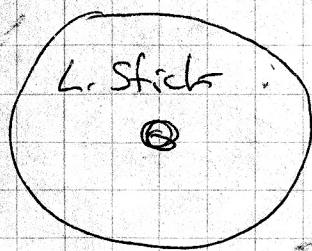
Bombloader	2
Ignitor (V1)	11
Inchworm	24
Little Arm	29
Running Machine	32
Screw Machine	41
V1	48

Bomb Loader Channel Assignments

Fet 5 
Fet 6 

By Gear Saw.

Same (FET 1)

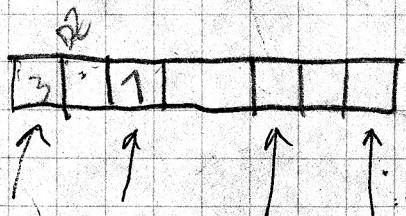
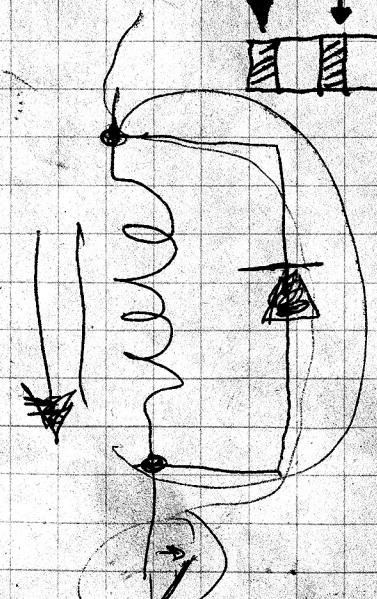
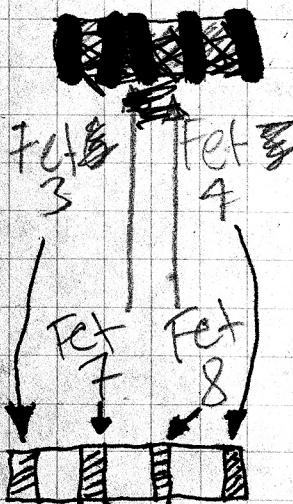


Same (FET 2)

R. stick



Same (C.V.)



Same
(C.V.)

Bomb Loader Update

March 1997
Ralf B.

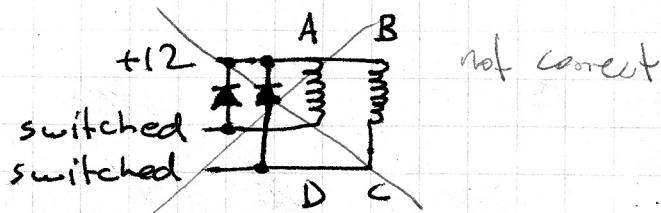
- relays now use 12V coils, so that one bank of FETs can be used for both relays and pegasus valves. See below
- 2nd 10 pin mil connector for new prop control lines
- remote start and stop disabled to free up channels for prop

10-pin connector Description: box wire: external wire:
Pin:

A	+ 12 V In	red/wh	
B	+ 12 V Out	purple	purple
C	FET 3	lt. blue/wh	lt. grey/wh
D	4	violet/wh	dk. grey/wh
E	5	green/wh	green/wh
F	6	black/wh	black/wh
G	N/C		
H	N/C		
J	FET 7	blue/wh	blue/wh
K	8	orange/wh	red/wh

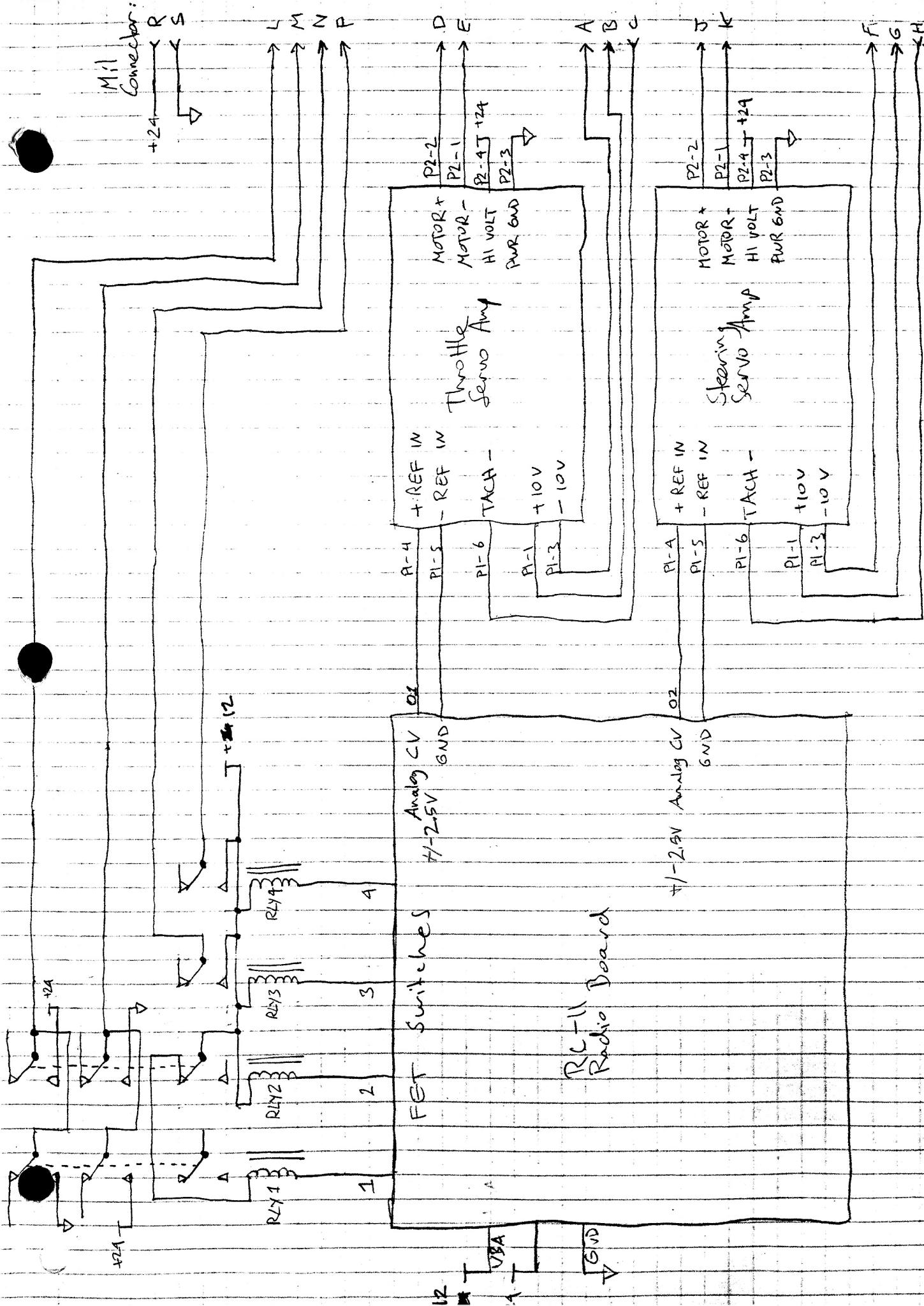
Pegasus valve connectors:

- A - Black
- B - Blue
- C - Green
- D - Brown



Sept. '97: FETs 1 and 2 now drive high-current relays with 24V coils. ~~The~~ The clamping diodes on the Fogarty board have been removed to drive both 12 and 24 VDC loads from the FET banks.

Bomb Loader Cart controller, Nov. 195



Bomb Loader Cart Controller

November 1995

This controller is like the one developed for the car hunt. One of Mike Fogarty's radio boards controls two high-current servo amplifiers, giving two proportional, closed-loop channels. Three additional channels are used to switch relays.

Mil connector pinouts:

A: Throttle, +10V ref	E: Throttle, -Motor	J: Steer, + Motor	N: Relay 3
B: ", -10V ref	F: Steer, +10V ref	K: ", -Motor	P: Relay 4
C: ", pot wiper	G: ", -10V ref	L: Arm Lift	R: +24 VDC
D: ", + Motor	H: ", pot wiper	M: Arm Lift	S: Ground

Servomotor mil connector pinouts:

- A: motor
- B: motor
- C: +10V ref
- D: pot wiper
- E: -10V ref

Servo amplifier switch settings:

- | | | | | |
|--------|--------|-------|--------|---------|
| 1: off | 2: off | 3: on | 4: on | 5: off |
| 6: on | 7: off | 8: on | 9: off | 10: off |

Joystick assignments:

Right stick, L-R: Steer L-R (proportional)

Right stick, U-D: Drive Fwd-Back (proportional)

Left stick, L-R: Full right turns on relay 4 = start engine big hand hydro cylinders

Left stick, U-D: Full travel up raises arm, down lowers. (Uses relays 1 and 2 to apply opposite polarity to actuator)

Gear switch: Turns on relay 3 = kill engine. Up is engine on.

big hand hydro
cylinders
gate open-close

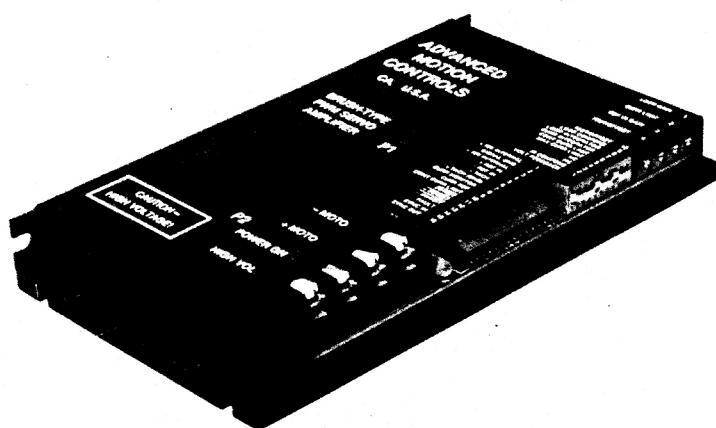
* On off switch on dash will override this
i.e. if the dashboard switch is on you
can't stop the engine remotely.

SERIES 50A PWM SERVO AMPLIFIERS
Models: 30A8, 25A20, 50A8, 50A20

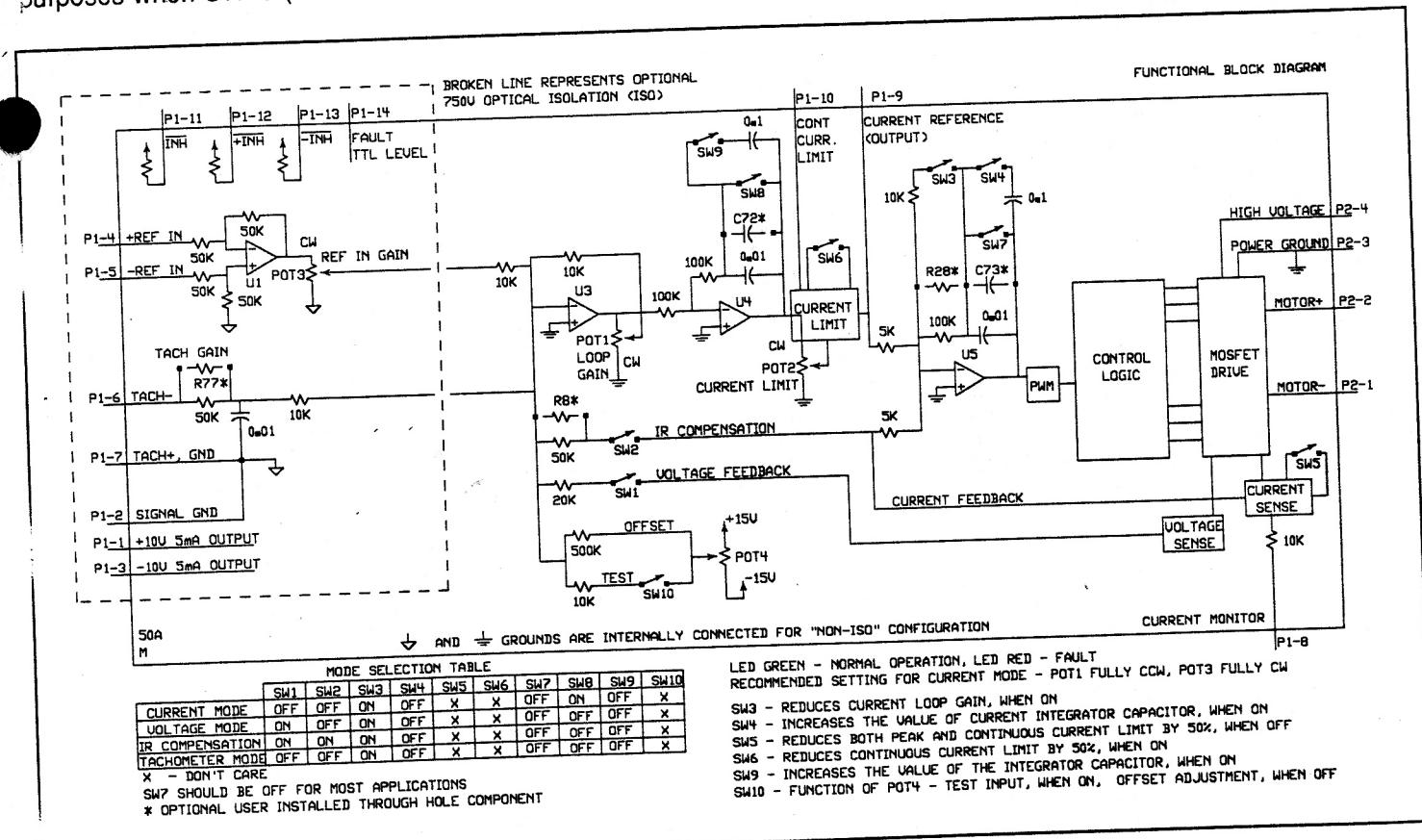
50A Series

FEATURES:

- * Surface-mount technology
- * Small size, low cost, ease of use
- * Optional input signal isolation for off-the-rectified 120 VAC line operation
- * DIP switch selectable: current, voltage, velocity, IR compensation, position loop control
- Four quadrant regenerative operation



DESCRIPTION: 50A Series PWM servo amplifiers are designed to drive brush type DC motors. Single red/green LED indicates operating status. All models are fully protected against over-voltage, over-current, over-heating and short-circuits across motor, ground and power leads. All models interface with digital controllers or can be used as a stand-alone drive. They require only a single unregulated DC power supply. Loop gain, current limit, input gain and offset can be adjusted using 15-turn potentiometers. The offset adjusting potentiometer can also be used as an on-board input signal for testing purposes when SW10 (DIP switch) is "On".



ADVANCED MOTION CONTROLS

3211 Corte Malpaso Unit 407, Camarillo, CA 93012

Tel: (805) 389-1935, Fax: (805) 389-1165

POWER STAGE SPECIFICATIONS		MODELS			
		30A8	25A20	50A8	50A20
DC SUPPLY VOLTAGE		20-80V	30-200V	20-80V	30-200V
PEAK CURRENT (2 sec. max., internally limited)		$\pm 30A$	$\pm 25A$	$\pm 50A$	$\pm 50A$
MAX. CONT. CURRENT (internally limited)		$\pm 15A$	$\pm 12.5A$	$\pm 25A$	$\pm 25A$
MINIMUM LOAD INDUCTANCE*		200uH	250uH	200uH	250uH
SWITCHING FREQUENCY		33kHz			
HEATSINK (BASE) TEMPERATURE RANGE		-25° to $+65^\circ$ C, shuts off if above $+65^\circ$ C			
POWER DISSIPATION AT CONT. CURRENT		30W	55W	50W	100W
OVER-VOLTAGE SHUT-DOWN (self reset)		86V	200V	86V	200V
BANDWIDTH		2.5kHz			

OUTPUT SPECIFICATIONS					
LOW VOLTAGE POWER SUPPLIES: (for customer use)		$+10V$ P1- 1 $-10V$ P1- 3	5mA, short circuit protected		
CURRENT MONITOR:	P1- 8		1V=4A, when SW5=On; 1V=2A, when SW5=Off	1V=8A, when SW5=On; 1V=4A, when SW5=Off	
CURRENT REFERENCE:	P1- 9		7.5V = maximum peak current		
FAULT:P1-14			TTL		

INPUT SPECIFICATIONS					
-INHIBIT:	P1-13		TTL, internal pull-up resistor, pull to ground to inhibit amplifier		
+INHIBIT:	P1-12				
INHIBIT:	P1-11				
TACHOMETER:	P1-6		Maximum $\pm 60V$ analog, 60K input resistance		
INPUT:	P1-4,5		Differential analog input, max. $\pm 15V$, 50K input resistance		

MECHANICAL SPECIFICATIONS					
POWER CONNECTOR		Screw terminals			
SIGNAL CONNECTOR		Molex connector			
SIZE (inches)		4.23 x 7.35 x 1.00			
WEIGHT		1.5 lb.			

* "Pancake" and "basket-wound" motors require external inductors

PIN FUNCTIONS

CONNECTOR	PIN	NAME	DESCRIPTION / NOTES	I/O
P1	1	+10V INTERNAL	Provides regulated voltages of +/-10V @ 5mA for customer use. Short circuit protected	O
	2	SIGNAL GND		
	3	-10V INTERNAL		
	4	+REF IN	Differential pre-amp	I
	5	-REF IN		
	6	-TACH	Tachometer	I
	7	+TACH (GND)		
	8	CURRENT MONITOR	This signal is proportional to the actual current in the motor leads. Scaling is 4A/V (2A/V when SW5=Off) for 30A8 and 25A20; and 8A/V (4A/V when SW5=Off) for 50A8 and 50A20	O
	9	CURRENT REF	Command signal to the internal current-loop. The maximum peak current rating of the amplifier always equals 7.5V at this pin (see section "D")	O
	10	CONTINUOUS CURRENT LIMIT	Can be used to reduce the factory-preset maximum continuous current limit (see section "D")	I
	11	<u>INHIBIT</u>	Inhibit. It turns off all mosfets of the "H" bridge drive when pulled to ground	I
	12	+ <u>INHIBIT</u>	Inhibits the motor for "+" direction only. This function can be useful to drive the motor off from a "limit switch"	I
	13	- <u>INHIBIT</u>	Inhibits the motor for "-" direction only. This function can be useful to drive the motor off from a "limit switch"	I
	14	FAULT (red LED)	TTL compatible output. It becomes high if output short-circuit, over-voltage, over-heating, inhibit, and during "power-on reset". Fault condition indicated by red LED	O
	15	SYNCH IN	Used for synchronizing the switching frequency of several amplifier modules. Consult factory for this option. Not applicable for "ISO" option.	N/A
	16	SYNCH OUT	In the "ISO" option pin 16 is connected to power ground and can be used as ground with P1- 8,9	

SWITCH FUNCTIONS

SWITCH	FUNCTION DESCRIPTION	SETTING	
		ON	OFF
1	Internal voltage feedback	On	Off
2	Internal current feedback for IR compensation	On	Off
3	Current loop gain	Decrease	Increase
4	Current loop integration	Increase	Decrease
5	Current scaling (when "Off", increases sensitivity of current sense thus reducing both peak and continuous current limit by 50% (see section "D"))		Half-current
6	Can be used to reduce factory-preset maximum continuous current limit (see section "D")		
7	It is recommended to leave SW7 in "Off" position	Shorts out the current loop integrator capacitor	Off
8	This capacitor normally ensures "error-free" operation by reducing the error-signal (output of summing amplifier) to zero	Shorts out the outer velocity / voltage loop integrator capacitor	Off
9	Adjusts the value of the integrator capacitor. It is recommended to leave SW9 in "Off" position for most applications	Increase	Decrease
10	Offset / test. Sensitivity of the "offset" pot. Used as an on-board reference signal in test mode	Increase	Decrease

POTENTIOMETER FUNCTIONS

POTENTIOMETER	DESCRIPTION	TURNING CW
Pot 1	Loop gain adjustment in voltage & velocity modes. Voltage to current scaling factor adjustment in current mode	Increases loop gain
Pot 2	Current limit. It adjusts both continuous and peak current limit by maintaining their ratio	Increases current limit
Pot 3	Reference gain. It adjusts the ratio between input signal and output variables (voltage, current, velocity)	Increases reference input gain
Pot 4	Offset / test. Used to adjust any imbalance in the input signal or in the amplifier. When SW10 (DIP switch) is "On", the sensitivity of this pot is greatly increased thus it can be used as an on-board signal source for testing purposes	N/A

8. CAUTIONARY NOTES

DO NOT REVERSE THE POWER SUPPLY LEADS !!!!

USE SUFFICIENT POWER SUPPLY CAPACITANCE !!!

Insufficient power supply capacitance problem occurs particularly with high inductance motors. During braking much of the stored mechanical energy is fed back into the power supply and charges its output capacitor to a higher voltage. If the charge reaches the amplifier's over-voltage shutdown point, output current and braking will cease. At that time energy stored in the motor inductance continues to flow through diodes in the amplifier to further charge the power supply capacitor. The voltage rise depends upon the power supply capacitance, motor speed, and inductance.

A 2mH motor at 20 amperes can charge 2000uF up to 30V.

An appropriate capacitance is typically 2000uF/A maximum output current for a 50V supply.

DO NOT SPIN THE MOTOR WITHOUT POWER !!!!

The motor acts as a generator and will charge up the power supply capacitors through the amplifier. Too high a speed may cause over-voltage breakdown in the power MOSfets. Note that an amplifier having an internal power converter that operates from the high voltage supply will become operative.

DO NOT SHORT THE MOTOR AT A HIGH SPEED !!!!

When the motor is shorted, its own generated voltage may produce a current flow as high as 10 times the amplifier peak current. The short itself should not damage the amplifier but may damage the motor. If the connection arcs or opens while the motor is spinning rapidly, this high current flows back into the amplifier (due to stored energy in the motor inductance) and may damage the amplifier.

ELECTRICAL CHARACTERISTICS — continued ($T_C = 25^\circ\text{C}$ unless otherwise noted.)

Characteristic	Symbol	Min	Typ	Max	Unit
ON CHARACTERISTICS (1)					
DC Current Gain ($I_C = 0.5 \text{ Adc}, V_{CE} = 5 \text{ Vdc}$) ($I_C = 4 \text{ Adc}, V_{CE} = 5 \text{ Vdc}$)	h_{FE}	50 200	100 400	—	—
Collector-Emitter Saturation Voltage ($I_C = 4 \text{ Adc}, I_B = 0.2 \text{ Adc}$) ($I_C = 8 \text{ Adc}, I_B = 0.4 \text{ Adc}$) ($I_C = 4 \text{ Adc}, I_B = 0.2 \text{ Adc}, T_C = 100^\circ\text{C}$)	$V_{CE(\text{sat})}$	— — —	— — —	2 3 2.2	Vdc
Base-Emitter Saturation Voltage ($I_C = 4 \text{ Adc}, I_B = 0.2 \text{ Adc}$) ($I_C = 8 \text{ Adc}, I_B = 0.4 \text{ Adc}$) ($I_C = 4 \text{ Adc}, I_B = 0.2 \text{ Adc}, T_C = 100^\circ\text{C}$)	$V_{BE(\text{sat})}$	— — —	— — —	2.5 3.5 2.4	Vdc
Diode Forward Voltage (2) ($I_F = 5 \text{ Adc}$)	V_f	—	—	2.5	Vdc

SWITCHING CHARACTERISTICS

Typical Resistive Load (Table 1)					
Delay Time	$(V_{CC} = 250 \text{ Vdc}, I_C(\text{pk}) = 6 \text{ A}$ $I_{B1} = I_{B2} = 0.25 \text{ A}, t_p = 25 \mu\text{s}$, Duty Cycle $\leq 1\%$)	t_d	—	0.04	μs
Rise Time		t_r	—	0.5	μs
Storage Time		t_s	—	8	μs
Fall Time		t_f	—	2	μs
Inductive Load, Clamped (Table 1)					
Voltage Storage Time	$(I_C(\text{pk}) = 6 \text{ A}, V_{CE(\text{pk})} = 250 \text{ Vdc}$ $I_{B1} = 0.06 \text{ A}, V_{BE(\text{off})} = 5 \text{ Vdc}$	t_{sv}	—	4	μs
Crossover Time		t_c	—	2	μs

(1) Pulse Test: Pulse Width = 300 μs , Duty Cycle = 2%.

(2) The internal Collector-to-Emitter diode can eliminate the need for an external diode to clamp inductive loads. Tests have shown that the Forward Recovery Voltage (V_f) of this diode is comparable to that of typical fast recovery rectifiers.

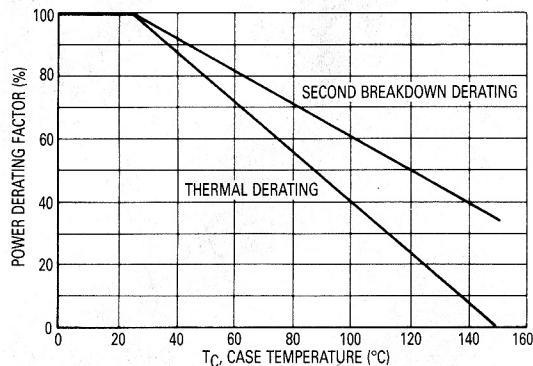
TYPICAL CHARACTERISTICS


Figure 1. Power Derating

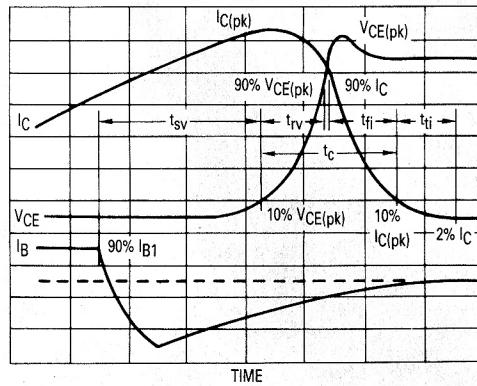


Figure 2. Inductive Switching Measurements

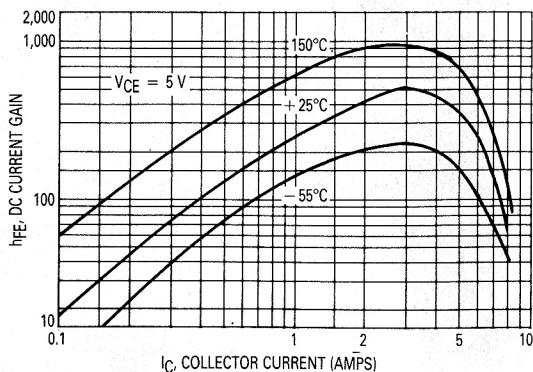


Figure 3. DC Current Gain

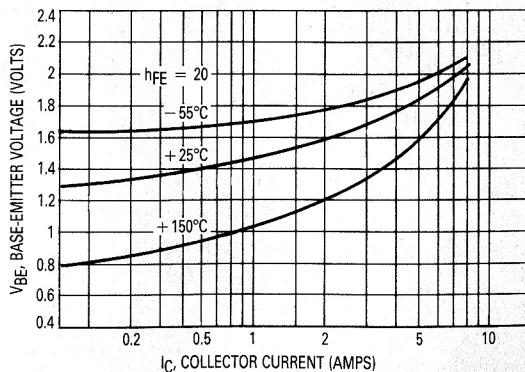


Figure 4. Base-Emitter Voltage

Table 1. Test Conditions for Dynamic Performance

Reverse Bias Safe Operating Area and Inductive Switching		Resistive Switching
Test Circuits	<p>NOTE PW and V_{CC} Adjusted for Desired I_C R_B Adjusted for Desired I_{B1}</p>	
Circuit Values	COIL DATA: FERROXCUBE CORE #6656 FULL BOBBIN (~16 TURNS) #16	GAP FOR 200 μ H/20 A $L_{coil} = 200 \mu$ H
Test Waveforms	<p>Output Waveforms</p> <p>t_1 ADJUSTED TO OBTAIN I_C</p> $t_1 \approx \frac{L_{coil} (I_{Cpk})}{V_{CC}}$ <p>$t_2 \approx \frac{L_{coil} (I_{Cpk})}{V_{clamp}}$</p>	$V_{CC} = 30$ V $V_{CE(pk)} = 250$ Vdc $I_C(pk) = 6$ A

OUTLINE DIMENSIONS

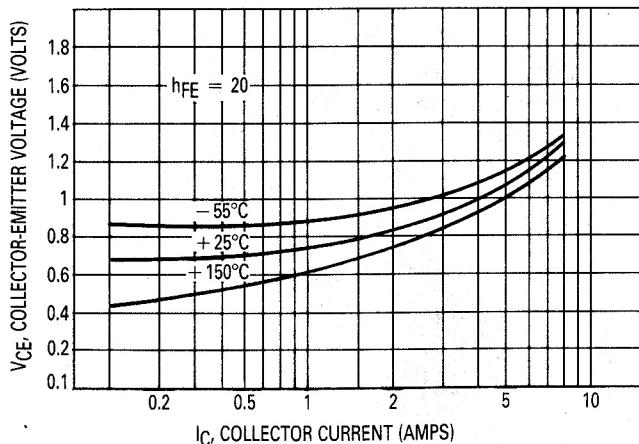
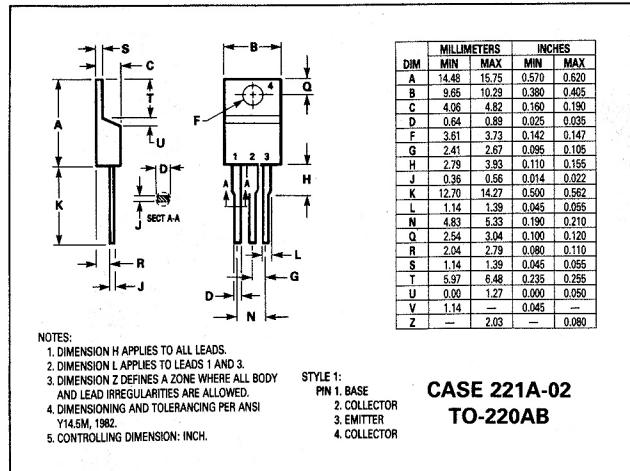


Figure 5. Collector Saturation Voltage



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SAFE OPERATING AREA INFORMATION

FORWARD BIAS

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate $I_C - V_{CE}$ limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figure 6 is based on $T_C = 25^\circ\text{C}$; $T_J(\text{pk})$ is variable depending on power level. Second breakdown pulse limits are valid for duty cycles to 10% but must be derated when $T_C \geq 25^\circ\text{C}$. Second breakdown limitations do not derate the same as thermal limitations. Allowable current at the voltages shown on Figure 6 may be found at any case temperature by using the appropriate curve on Figure 1.

REVERSE BIAS

For inductive loads, high voltage and high current must be sustained simultaneously during turn-off, in most cases, with the base to emitter junction reverse biased. Under these conditions the collector voltage must be held to a safe level at or below a specific value of collector current. This can be accomplished by several means such as active clamping, RC snubbing, load line shaping, etc. The safe level for these devices is specified as Reverse Bias Safe Operating Area and represents the voltage-current condition allowable during reverse biased turn-off. This rating is verified under clamped conditions so that the device is never subjected to an avalanche mode. Figure 7 gives the complete RBSOA characteristics.

The Safe Operating Area figures shown in Figures 6 and 7 are specified ratings for these devices under the test conditions shown.

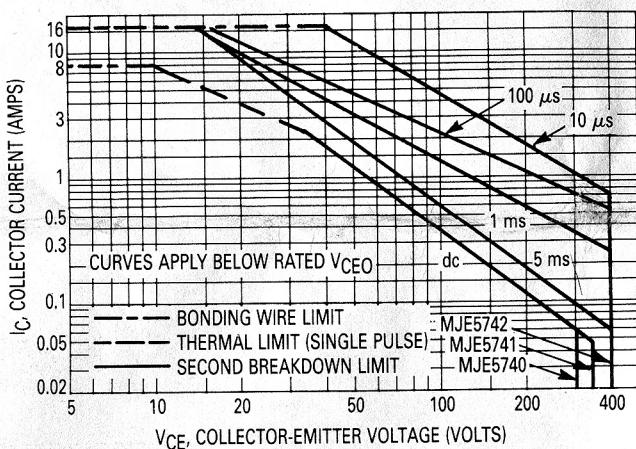


Figure 6. Forward Bias Safe Operating Area

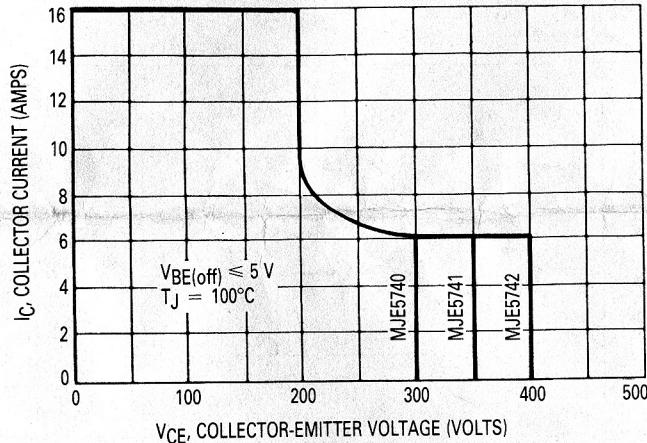


Figure 7. Reverse Bias Safe Operating Area

RESISTIVE SWITCHING PERFORMANCE

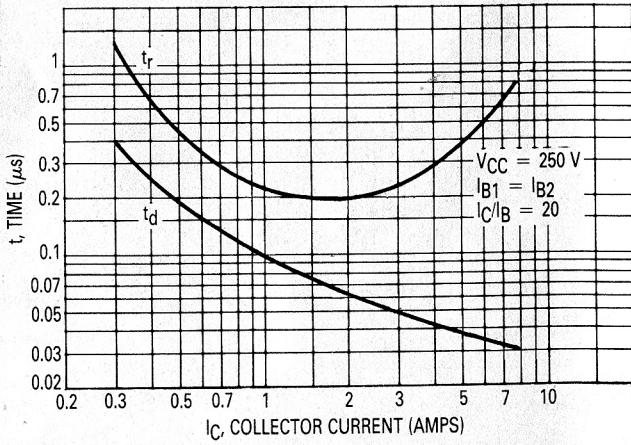


Figure 8. Turn-On Time

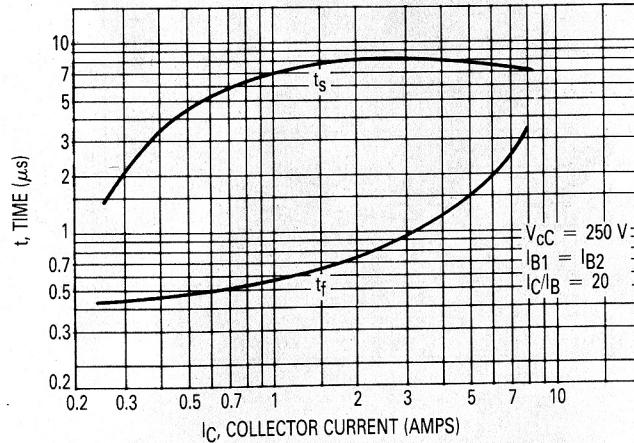


Figure 9. Turn-Off Time

Literature Distribution Centers:

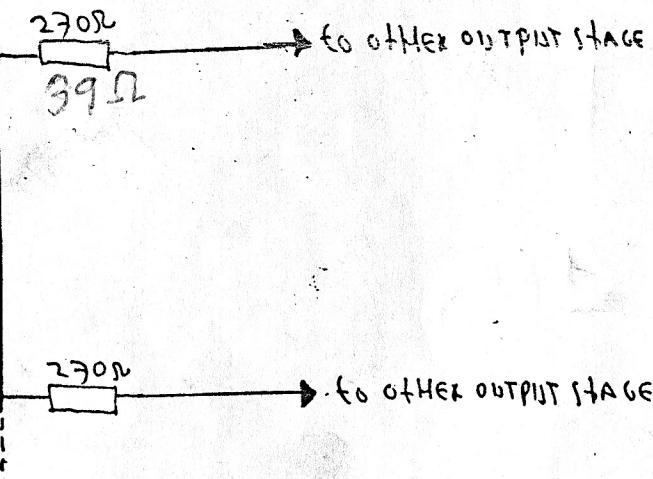
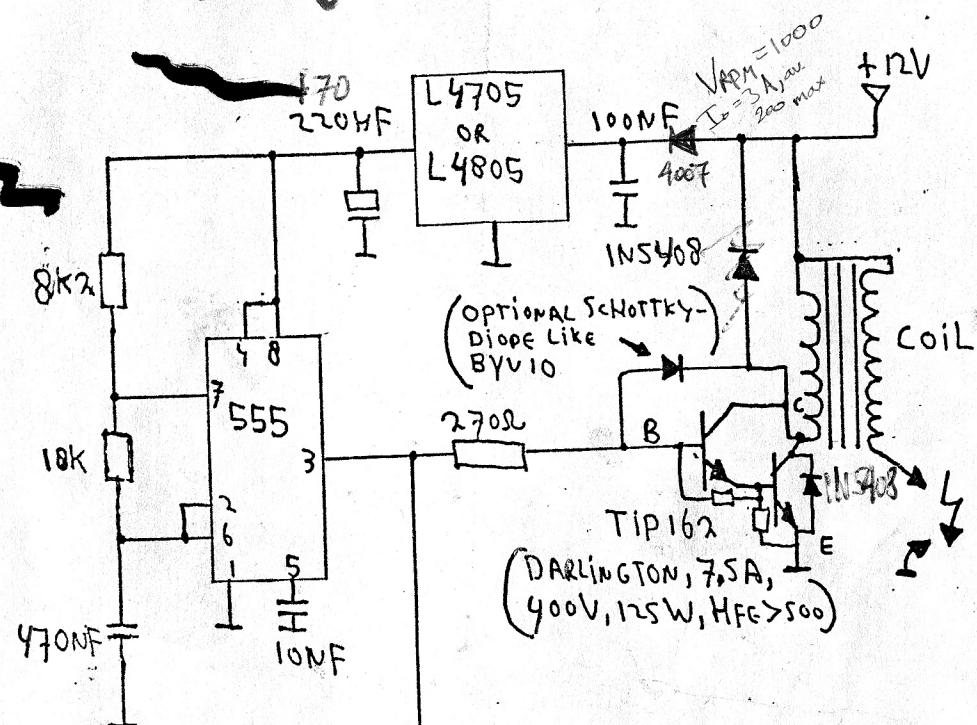
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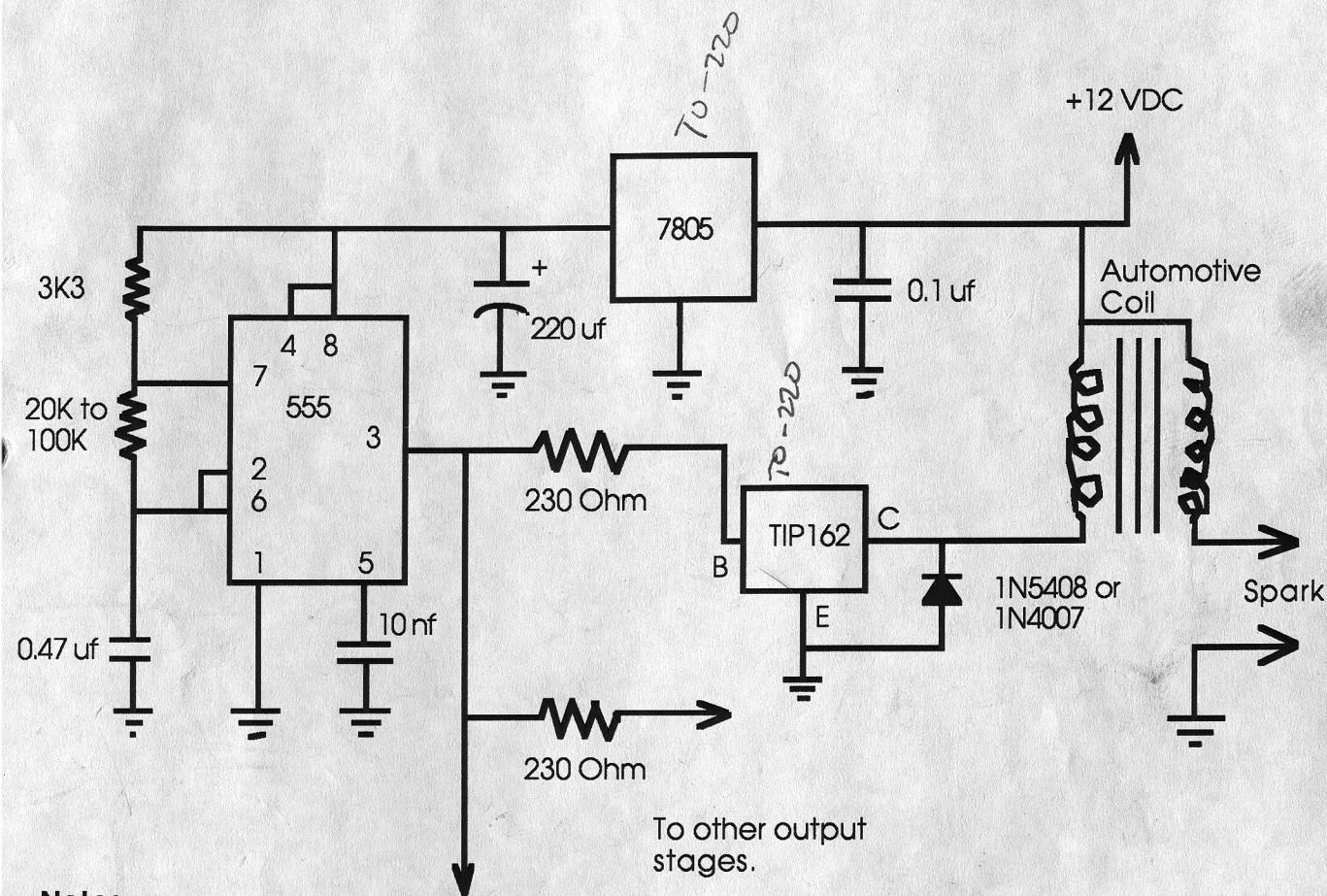
MOTOROLA





Electronic Ignition Circuit Survival Research Laboratories

Revision Feb. 1992 by Ralf



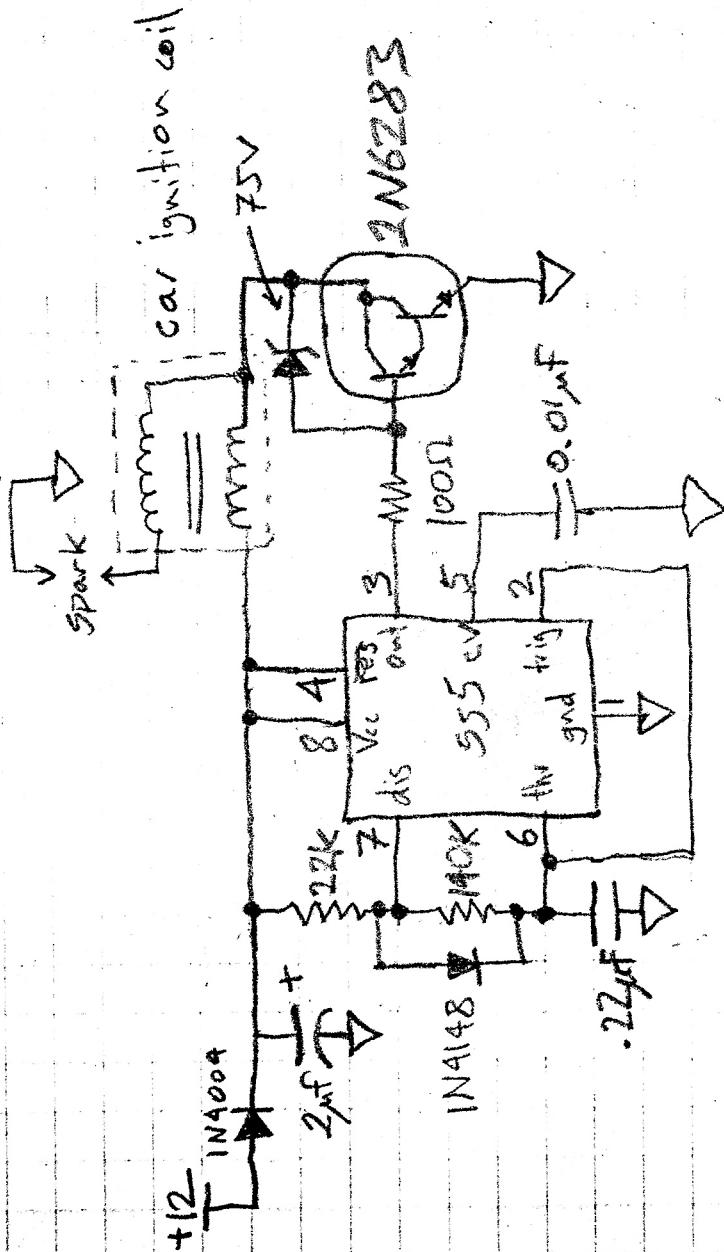
Notes:

1. The letter "u" in capacitor values denotes micro.
 2. The TIP162 is a Darlington transistor package, and provides good results when driven by any 50% square waveform from 14 to 150 Hz.

Apr. 92 : 2 of these circuits now in V1, one of which drives 2 output stages.

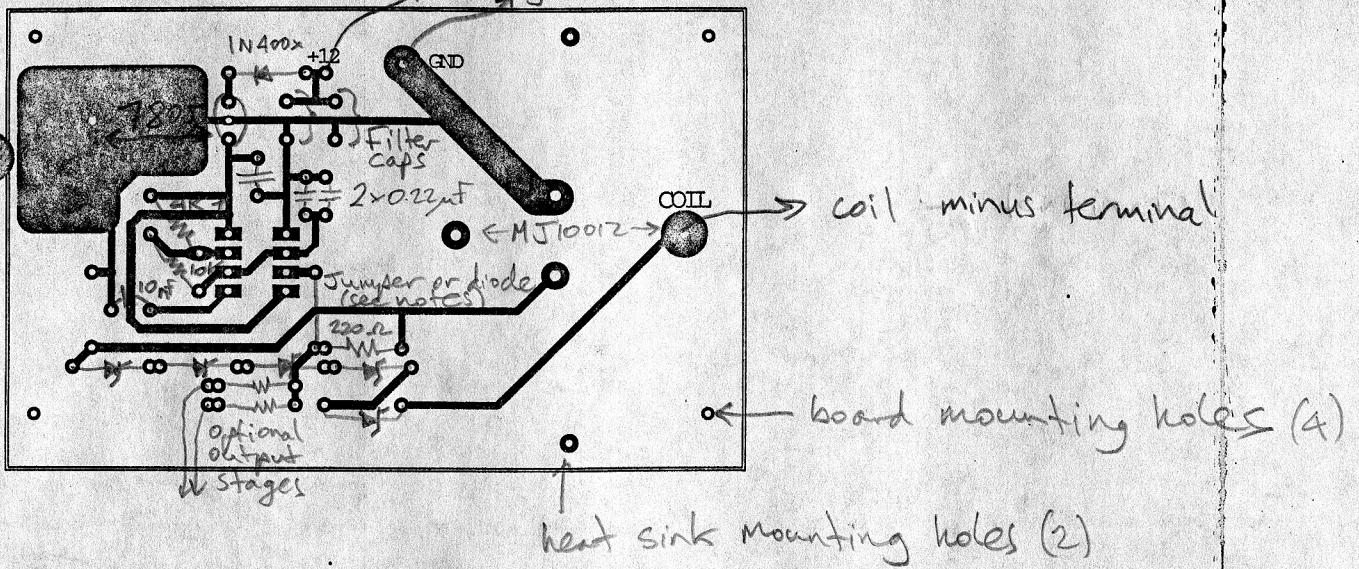
Apr. 94: Newer version of this circuit available under Ignitor section.

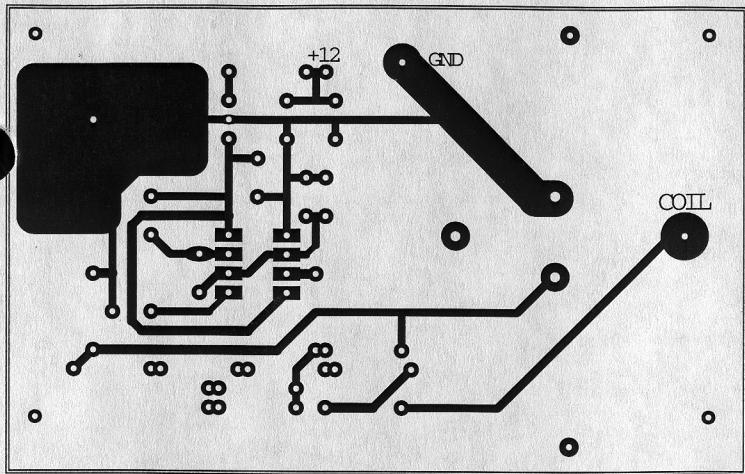
4 of these have been built to make use of some Darlington transistors we got free. The low breakdown voltage (75v) means they'll only make a quarter inch or so of spark. Note the diode to give a low (~20%) duty cycle and a lower sparking frequency. Other than the V₁-type ignitor, the printed circuit board layout file is "Small TO-3 ignition".



Ralf B.
March 1996

Parts Placement







MOTOROLA

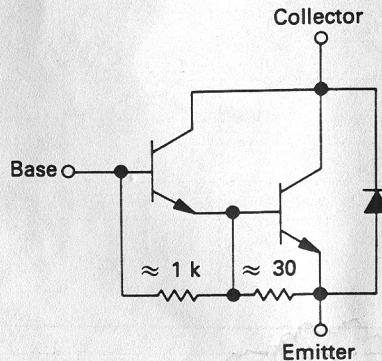
SEMICONDUCTORS

P.O. BOX 20912 • PHOENIX, ARIZONA 85036

NPN SILICON POWER DARLINGTON TRANSISTOR

The MJ10012 and MJH10012 are high-voltage, high-current darlington transistors designed for automotive ignition, switching regulator and motor control applications.

- Collector-Emitter Sustaining Voltage —
 $V_{CEO(sus)} = 400$ Vdc (Min)
- 175 Watts Capability at 50 Volts
- Automotive Functional Tests



MAXIMUM RATINGS

Rating	Symbol	MJ10012	MJH10012	Unit
Collector-Emitter Voltage	$V_{CEO(sus)}$	400		Vdc
Collector-Emitter Voltage ($R_{BE} = 27 \Omega$)	V_{CER}	550		Vdc
Collector-Base Voltage	V_{CBO}	600		Vdc
Emitter-Base Voltage	V_{EBO}	8.0		Vdc
Collector Current — Continuous — Peak (1)	I_C	10 15		Adc
Base Current	I_B	2.0		Adc
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ @ $T_C = 100^\circ\text{C}$ Derate above 25°C	P_D	175 100 1.0	118 47.5 1.05	Watts Watts W/W°C
Operating and Storage Junction Temperature Range	T_J, T_{stg}	-65 to +200	-55 to +150	°C

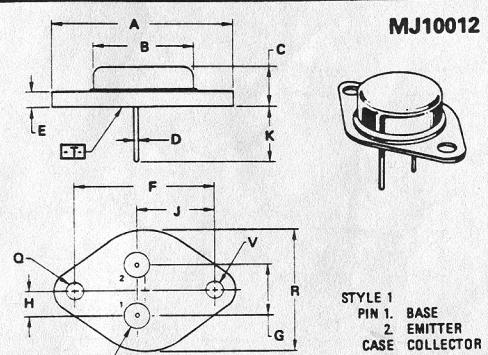
THERMAL CHARACTERISTICS

Characteristic	Symbol	Max		Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	1.0	0.95	°C/W
Maximum Lead Temperature for Soldering Purposes: 1/8" from Case for 5 Seconds	T_L	275		°C

(1) Pulse Test: Pulse Width = 5.0 ms, Duty Cycle ≤ 10%.

MJ10012
MJH10012

10 AMPERE

POWER TRANSISTORS
DARLINGTON NPN SILICON400 VOLTS
175 AND 118 WATTSMJ10012
STYLE 1
PIN 1. BASE
2. Emitter
CASE COLLECTOR

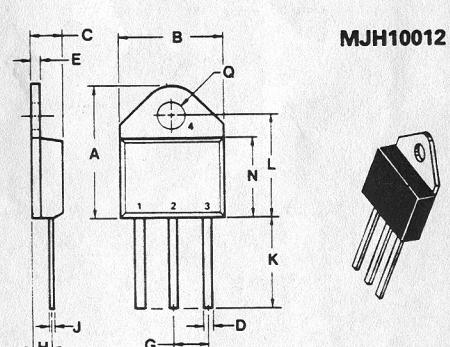
- NOTES:
 1. DIMENSIONS Q AND V ARE DATUMS.
 2. T IS SEATING PLANE AND DATUM.
 3. POSITIONAL TOLERANCE FOR
MOUNTING HOLE D:
 $\Delta \pm 0.13 (0.005) \odot$ T V Q
 FOR LEADS:
 $\Delta \pm 0.13 (0.005) \odot$ T V Q O
 4. DIMENSIONS AND TOLERANCES PER
ANSI Y14.5, 1973.

CASE 1-05
TO-204AA
(Formerly TO-3)

MILLIMETERS		INCHES		
DIM	MIN	MAX	MIN	MAX
A	—	39.37	—	1.550
B	—	21.08	—	0.830
C	6.35	7.52	0.250	0.300
D	0.97	1.09	0.038	0.043
E	1.40	1.78	0.055	0.070
F	30.15	30.50	1.187	1.200
G	10.92	11.25	0.430	0.450
H	5.46	5.82	0.215	0.230
J	16.89	17.25	0.665	0.680
K	11.18	12.19	0.440	0.480
L	3.81	4.19	0.151	0.165
R	—	26.67	—	1.050
U	4.83	5.33	0.190	0.210
V	3.81	4.19	0.151	0.165

CASE 1-05
TO-204AA

(Formerly TO-3)



MJH10012

- STYLE 1:
PIN 1. BASE
2. Emitter
3. Collector

MILLIMETERS		INCHES		
DIM	MIN	MAX	MIN	MAX
A	20.32	21.08	0.800	0.830
B	15.49	15.90	0.610	0.626
C	4.19	5.08	0.165	0.200
D	1.02	1.65	0.040	0.065
E	1.35	1.65	0.053	0.065
G	5.21	5.72	0.205	0.225
H	2.41	3.20	0.095	0.126
J	0.38	0.64	0.015	0.025
K	12.70	15.49	0.500	0.610
L	15.88	16.51	0.625	0.650
M	12.19	12.70	0.480	0.500
Q	4.04	4.22	0.159	0.166

CASE 340-01
TO-218AC

ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS (1)					
Collector-Emitter Sustaining Voltage (Figure 1) ($I_C = 200 \text{ mA}_\text{dc}$, $I_B = 0$, $V_{\text{clamp}} = \text{Rated } V_{\text{CEO}}$)	$V_{\text{CEO(sus)}}$	400	—	—	Vdc
Collector-Emitter Sustaining Voltage (Figure 1) ($I_C = 200 \text{ mA}_\text{dc}$, $R_{BE} = 27 \text{ Ohms}$, $V_{\text{clamp}} = \text{Rated } V_{\text{CER}}$)	$V_{\text{CER(sus)}}$	425	—	—	Vdc
Collector Cutoff Current (Rated V_{CER} , $R_{BE} = 27 \text{ Ohms}$)	I_{CER}	—	—	1.0	mA_dc
Collector Cutoff Current (Rated V_{CBO} , $I_E = 0$)	I_{CBO}	—	—	1.0	mA_dc
Emitter Cutoff Current ($V_{EB} = 6.0 \text{ Vdc}$, $I_C = 0$)	I_{EBO}	—	—	40	mA_dc
ON CHARACTERISTICS (1)					
DC Current Gain ($I_C = 3.0 \text{ Adc}$, $V_{CE} = 6.0 \text{ Vdc}$) ($I_C = 6.0 \text{ Adc}$, $V_{CE} = 6.0 \text{ Vdc}$) ($I_C = 10 \text{ Adc}$, $V_{CE} = 6.0 \text{ Vdc}$)	h_{FE}	300 100 20	550 350 150	— 2000 —	—
Collector-Emitter Saturation Voltage ($I_C = 3.0 \text{ Adc}$, $I_B = 0.6 \text{ Adc}$) ($I_C = 6.0 \text{ Adc}$, $I_B = 0.6 \text{ Adc}$) ($I_C = 10 \text{ Adc}$, $I_B = 2.0 \text{ Adc}$)	$V_{CE(\text{sat})}$	— — —	— — —	1.5 2.0 2.5	Vdc
Base-Emitter Saturation Voltage ($I_C = 6.0 \text{ Adc}$, $I_B = 0.6 \text{ Adc}$) ($I_C = 10 \text{ Adc}$, $I_B = 2.0 \text{ Adc}$)	$V_{BE(\text{sat})}$	— —	— —	2.5 3.0	Vdc
Base-Emitter On Voltage ($I_C = 10 \text{ Adc}$, $V_{CE} = 6.0 \text{ Vdc}$)	$V_{BE(\text{on})}$	—	—	2.8	Vdc
Diode Forward Voltage ($I_F = 10 \text{ Adc}$)	V_f	—	2.0	3.5	Vdc
DYNAMIC CHARACTERISTICS					
Output Capacitance ($V_{CB} = 10 \text{ Vdc}$, $I_E = 0$, $f_{\text{test}} = 100 \text{ kHz}$)	C_{ob}	—	165	350	pF
SWITCHING CHARACTERISTICS					
Storage Time ($V_{CC} = 12 \text{ Vdc}$, $I_C = 6.0 \text{ Adc}$, $I_{B1} = I_{B2} = 0.3 \text{ Adc}$) Figure 2	t_s	—	7.5	15	μs
Fall Time	t_f	—	5.2	15	μs
FUNCTIONAL TESTS					
Second Breakdown Collector Current with Base-Forward Biased	$I_{S/B}$	See Figure 10			—
Pulsed Energy Test (See Figure 12)	$\frac{I_{C2L}}{2}$	—	—	180	mJ

(1) Pulse Test: Pulse Width = 300 μs , Duty Cycle = 2%.

FIGURE 1 – SUSTAINING VOLTAGE TEST CIRCUIT

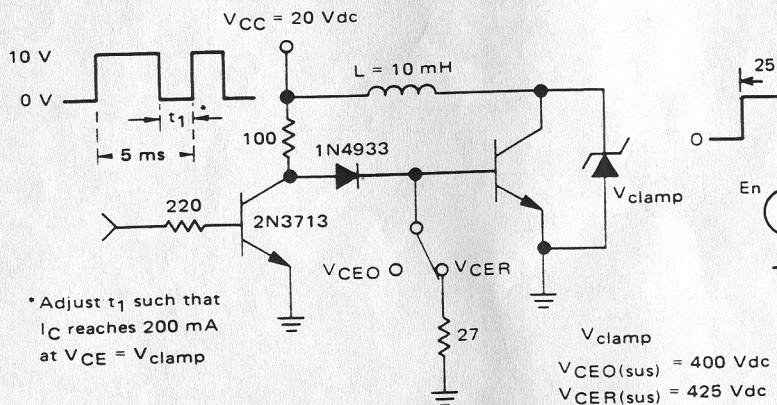
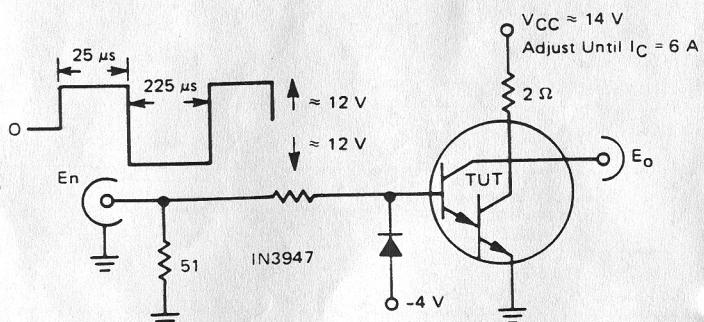


FIGURE 2 – SWITCHING TIMES TEST CIRCUIT



MOTOROLA Semiconductor Products Inc.

FIGURE 3 – DC CURRENT GAIN

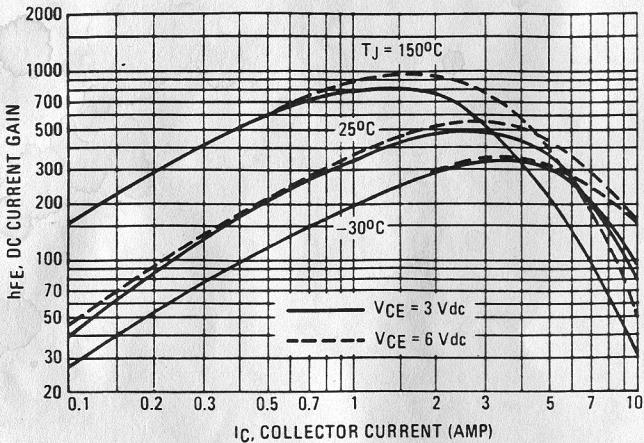


FIGURE 4 – COLLECTOR-SATURATION REGION

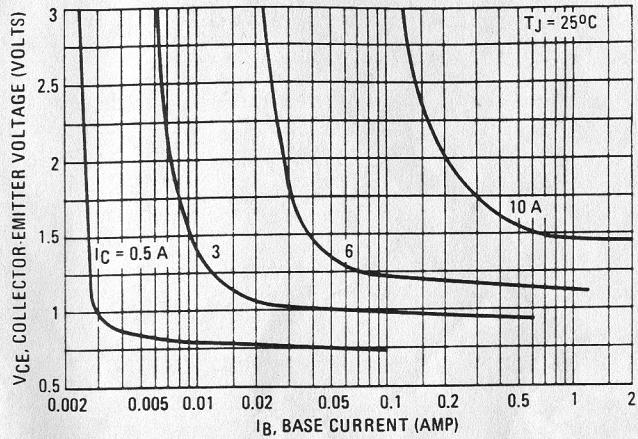


FIGURE 5 – COLLECTOR-EMITTER SATURATION VOLTAGE

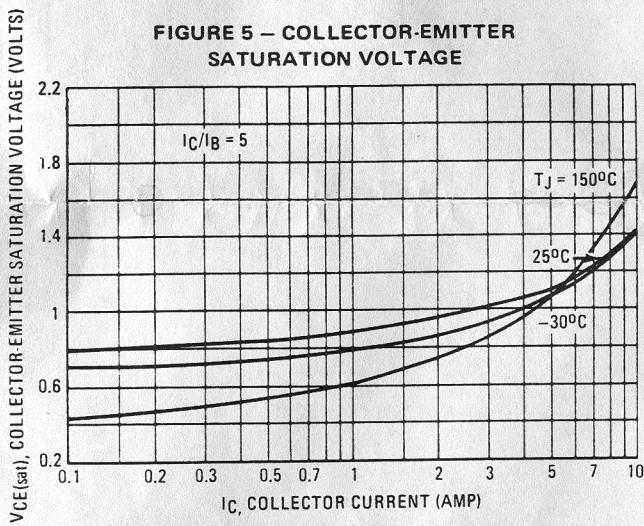


FIGURE 6 – BASE-EMITTER VOLTAGE

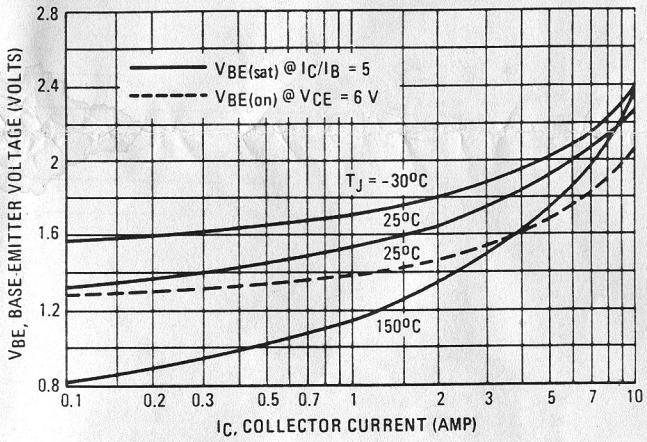


FIGURE 7 – TURN-OFF SWITCHING TIME

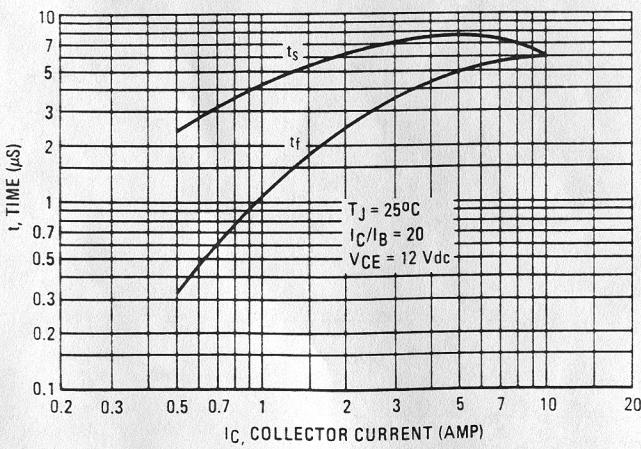
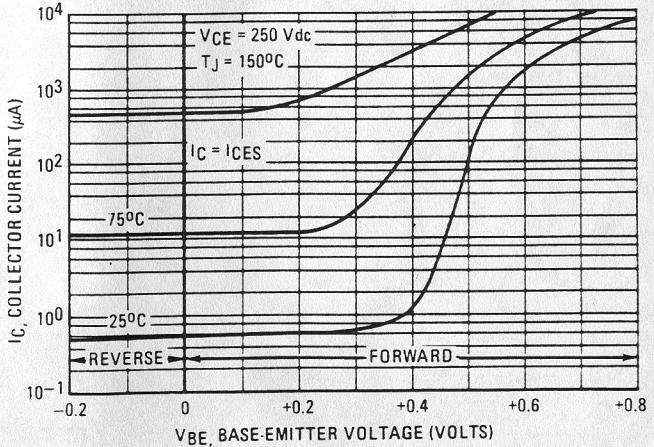
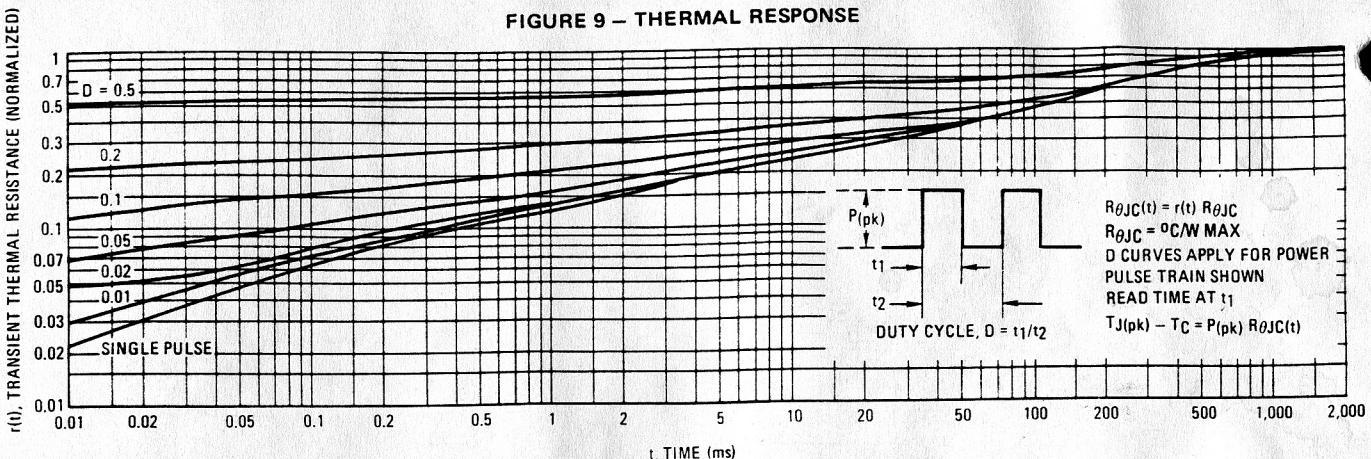
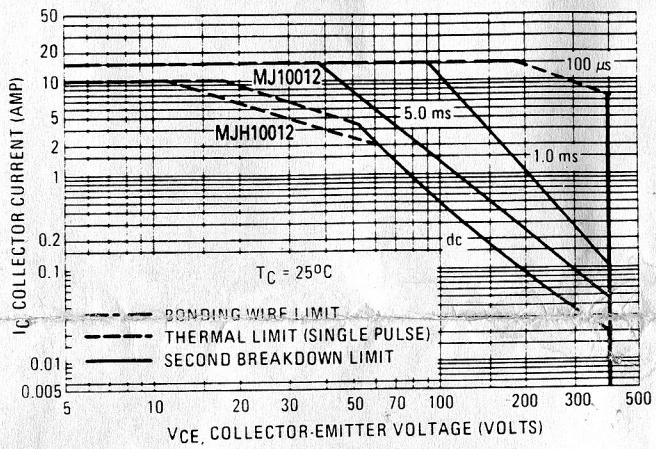
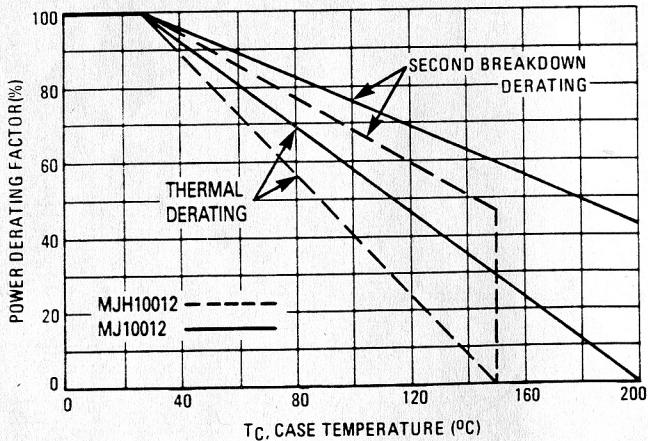


FIGURE 8 – COLLECTOR CUTOFF REGION

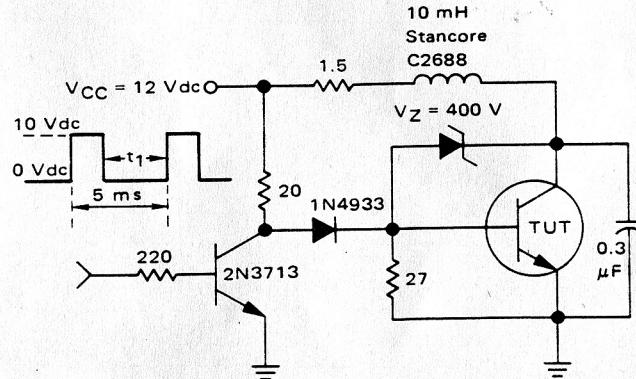


**FIGURE 10 – FORWARD BIAS SAFE OPERATING AREA****FIGURE 11 – POWER DERATING**

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate I_C - V_{CE} limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figure 10 is based on $T_C = 25^{\circ}\text{C}$; $TJ(pk)$ is variable depending on power level. Second breakdown pulse limits are valid for duty cycles to 10% but must be derated when $T_C \geq 25^{\circ}\text{C}$. Second breakdown limitations do not derate the same as thermal limitations. Allowable current at the voltages shown on Figure 10 may be found at any case temperature by using the appropriate curve on Figure 11.

$TJ(pk)$ may be calculated from the data in Figure 11. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

FIGURE 12 – USAGE TEST CIRCUIT

t_1 to be selected such that I_C reaches 6 Adc before switch-off.

NOTE:

"Usage Test," Figure 12 specifies energy handling capabilities in an automotive ignition circuit.

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MOTOROLA Semiconductor Products Inc.

BOX 20912 • PHOENIX, ARIZONA 85036 • A SUBSIDIARY OF MOTOROLA INC.

DS3318R2

NPN Silicon Power Darlington Transistors

The MJE5740, 41, 42 darlington transistors are designed for high-voltage power switching in inductive circuits. They are particularly suited for operation in applications such as:

- Small Engine Ignition
- Switching Regulators
- Inverters
- Solenoid and Relay Drivers
- Motor Controls

MAXIMUM RATINGS

Rating	Symbol	MJE5740	MJE5741	MJE5742	Unit
Collector-Emitter Voltage	$V_{CEO(sus)}$	300	350	400	Vdc
Collector-Emitter Voltage	V_{CEV}	600	700	800	Vdc
Emitter Base Voltage	V_{EB}		8		Vdc
Collector Current — Continuous — Peak (1)	I_C I_{CM}		8 16		Adc
Base Current — Continuous — Peak (1)	I_B I_{BM}		2.5 5		Adc
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above 25°C	P_D		2 16		Watts mW/ $^\circ\text{C}$
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ Derate above 25°C	P_D		80 640		Watts mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	T_J, T_{stg}		−65 to +150		°C

(1) Pulse Test: Pulse Width = 5 ms, Duty Cycle = 10%.

THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	1.56	°C/W
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	62.5	°C/W
Maximum Lead Temperature for Soldering Purposes: 1/8" from Case for 5 Seconds	T_L	275	°C

ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted.)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS (2)					
Collector-Emitter Sustaining Voltage ($I_C = 50 \text{ mA}$, $I_B = 0$)	$V_{CEO(sus)}$	300 350 400	— — —	— — —	Vdc
Collector Cutoff Current ($V_{CEV} = \text{Rated Value}$, $V_{BE(\text{off})} = 1.5 \text{ Vdc}$) ($V_{CEV} = \text{Rated Value}$, $V_{BE(\text{off})} = 1.5 \text{ Vdc}$, $T_C = 100^\circ\text{C}$)	I_{CEV}	— —	— 5	1	mAdc
Emitter Cutoff Current ($V_{EB} = 8 \text{ Vdc}$, $I_C = 0$)	I_{EBO}	—	—	75	mAdc

SECOND BREAKDOWN

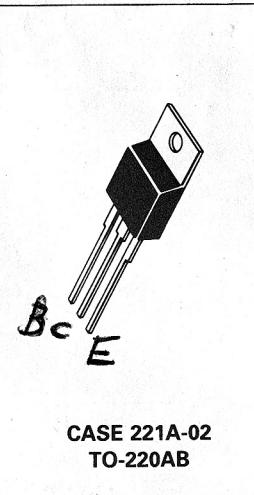
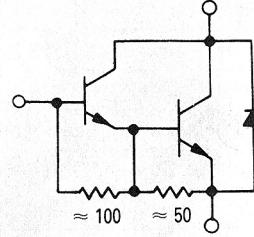
Second Breakdown Collector Current with Base Forward Biased	$I_{S/b}$	See Figure 6
Clamped Inductive SOA with Base Reverse Biased	R_{BSOA}	See Figure 7

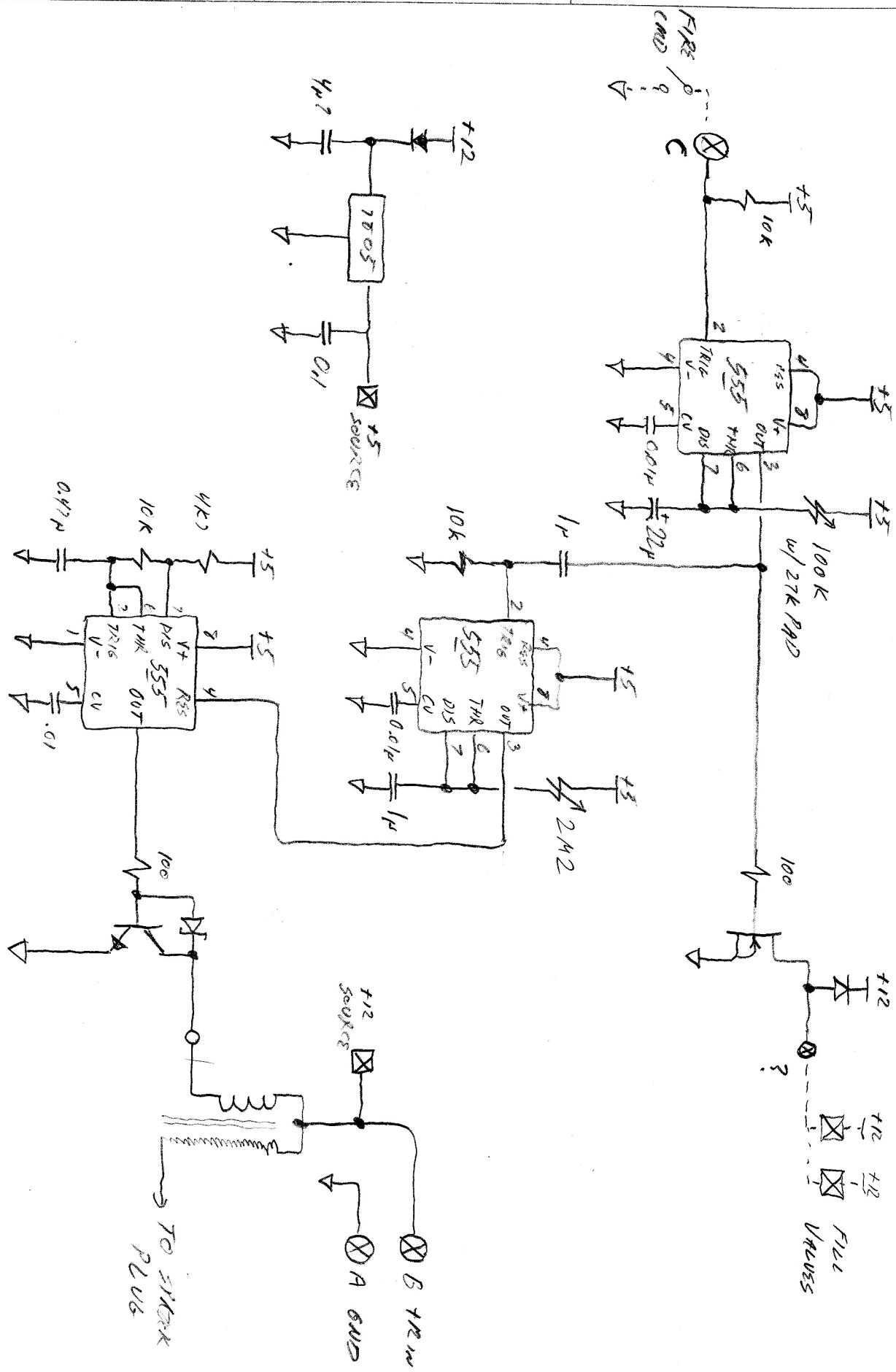
(2) Pulse Test: Pulse Width = 300 μs , Duty Cycle = 2%.

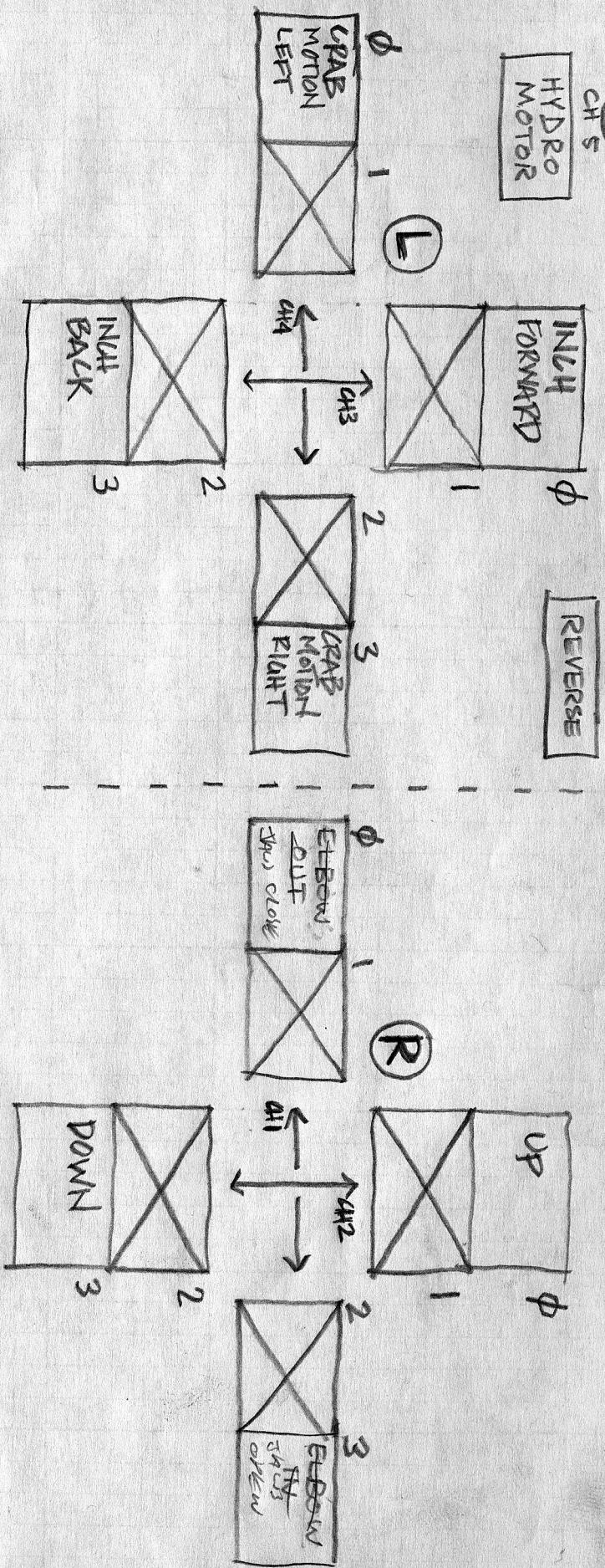
(continued)

MJE5740
MJE5741
MJE5742

**POWER DARLINGTON
TRANSISTORS**
8 AMPERES
300, 350, 400 VOLTS
80 WATTS



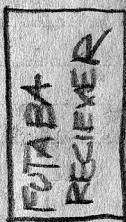


INCHWORM RC DIAGRAM
 REV 1 FWD


AUG 27 1992

 Mod 10/16/92
 G. Homsey

COMMAND	JOY POS	RELAY LEFT	RELAY RIGHT	JOY POS	COMMAND
FORWARD RED	CHG 3 (inverted)	C22	C11	CH2 Ø	UP BLACK STRIPE
HYDRO MOTOR ORG	CH5 3	A22	A11	CH1 Ø	ELBOW IN OUT YEL
REVERSE RED (x2)	CHG Ø	C19	A14	CH1 3	ELBOW IN YEL/BK
INCH FORWARD	CH3 Ø	A15	C14	CH2 3	DOWN BLACK
INCH BACK	CH3 3	A18	C15	CH4 Ø	CRAB LEFT BLUE STRIPE
CRAB RIGHT BLUE	CH4 3	C18		OR GATE BOARD	DUMP RELAY MAUENTA STRIPE



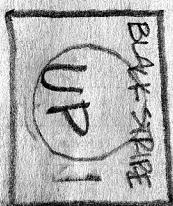
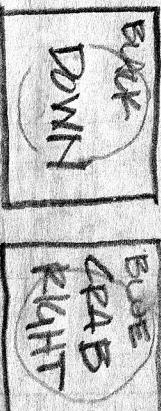
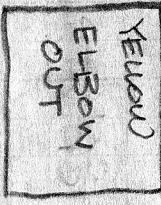
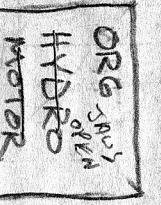
Aug 27 1992

Chay

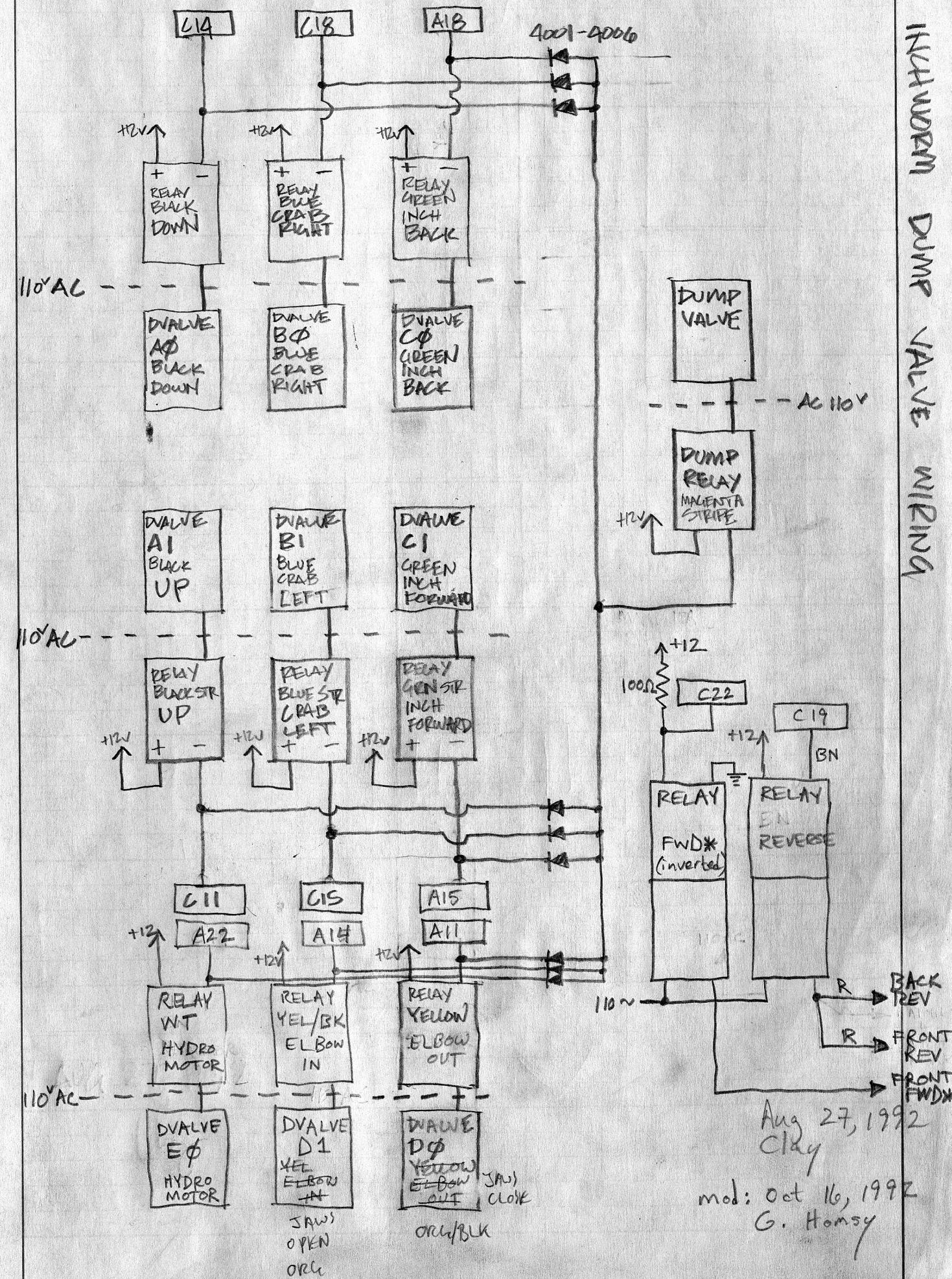
Mod 10/16/92
G. Homsy

INNORN HYDRO VALVES (DIRECTIONALS)

DIRECTIONAL VALVES

(A)  BLACK	(B)  BLUE	(C)  GREEN	(D)  YELLOW	(E)  ORANGE
--	--	--	---	--

(CENTER
 OR MIDDLE
 HENCE)



ED #0 - NO TAPE
ED #1 - TAPE

LITTLE ARM CONNECTOR
1 OF 2

SPECIAL INSTRUCTIONS

HITE #2 -
2 PIECES OF TAPE

1. RED #1

BK FG
RD GH
WH AJ

POWER JK WH POWER TB 1/2

RD/WH V

2. WHITE #0

PR/WH A
RD/WH B
BK/WH C

WHITE #1

PR/WH D
RED/WH E
BK/WH F

3. WHITE #2

PR/WH K
RD/WH L
BK/WH M

LITTLE ARM CONNECTOR
PAGE 1 OF 2

SPECIAL INSTRUCTIONS

1. RED #0

BK GN
RD P
WH R

2.

3.

Din 10 Pin mil:

A	A	2	4
B	C	2	4
C	A	2	2
D	C	2	2
E	A	2	0
F	C	2	0
G	C	1	8
H	A	1	8
J	C	1	6
K	A	1	6

A	A	1	2
B	A	1	1
C	C	1	1
D	A	1	0
E	A	9	
F	C	9	
G	A	8	
H	A	7	
J	C	7	
K	A	6	
L	A	5	
M	C	5	
N	A	4	
P	A	3	
R	C	3	

NO 5

NO 4

T to White wire
V to Black wire

Little Arm
Harness Pinouts
3/97

(A) clip black
to orange exposed wire

* 1 is shortest
5 is longest

1
A Pin K
B Pin J
C Pin K
D Pin J

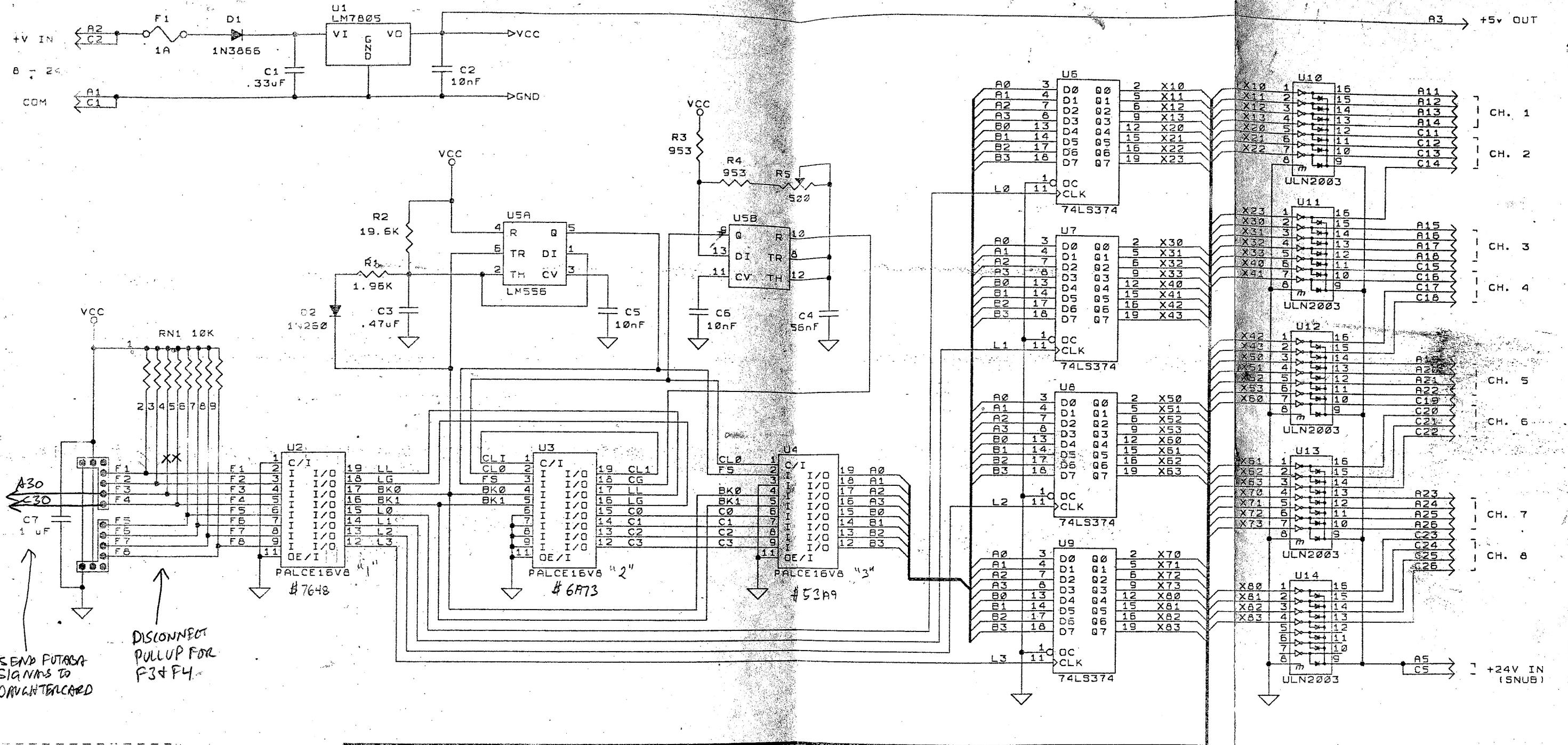
2
A Pin H
B Pin G
C Pin H
D Pin G

3
A Pin F
B Pin E
C Pin F
D Pin E

4
A Pin D
B Pin C
C Pin D
D Pin C

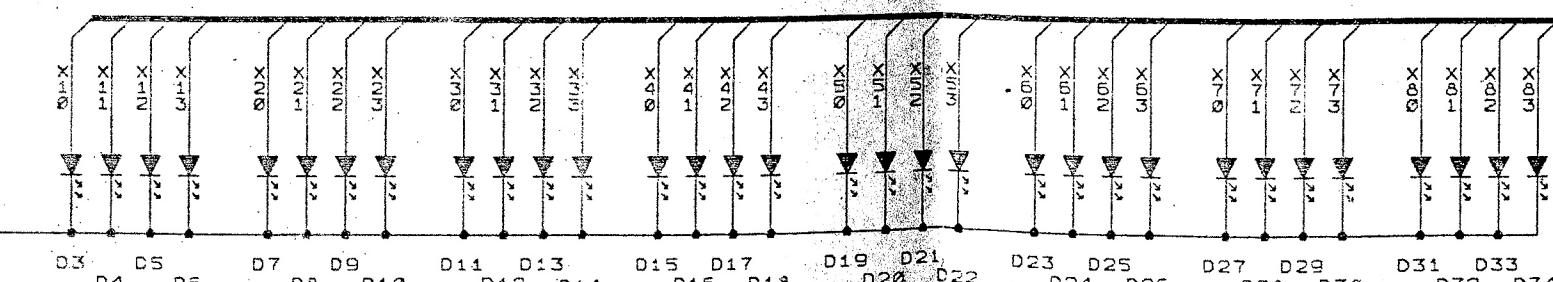
5
A Pin B
B - Pin A
C Pin B
D - Pin A

Little Arm
Harness
March 97



POWER NET

CHIP	VCC	GND	CAP
U2	20	10	C8
U3	20	10	C9
U4	20	10	C10
U5	14	7	C11
U6	20	10	C12
U7	20	10	C13
U8	20	10	C14
U9	20	10	C15
			C15 = .1 uF



D3-D34 ARE HLMP-1640

5v GREEN LED'S
WITH INTEGRAL RESISTOR

RUNNING MACHINE RC

WINGNUT METATRONICS

Paul Roybal
P.O. BOX 7128
MENLO PARK, CA. 94026



TITLE	8 CHANNEL PWM DECODER	
SIZE	Document Number	REV
B	BRL-PR-001-03 SCH	0
DATE	October 24, 1990 Sheet 1 of 1	

Running Machine Futata Transmitter

Mil Connector Pinouts

A +V

B Gnd

C Right Stick - L/R pot wiper



Jumper these
pins to restore
normal operation

D " cct board



E Right Stick - U/D pot wiper



F " cct board



G Left Stick - U/D pot wiper



H " cct board



J Left Stick - L/R pot wiper



K " cct board



L Gear Switch (switch to gnd)



M Gear Swict cct board lead

This connector allows use of external potentiometers.

Jan. 97

Ralf B.

Running machine 11 pin mil connector:

- A Steering Feedback pot or Red Ground
- B " Green +12
- C " Gray Black Black }
- D " Shield Green pot
- E " Red }
- F " Shield }
- G
- H
- I
- J
- K + 12 V from battery
- L Battery grd.

5-15-95 M. Fogarty

VALVE 2

D
C
B
A

stick R
TR
TR

CONTROL BOX CONN PIN

E
F
D
D

Red-A	Running machine control box:		
Green-B	-		
White-C	19 pin mil connector:		
Black-D	A	Value 1	$D \oplus B$
	B		A
	C		C
D		Value 2	$D \oplus B$
E			A + B
F			D
G		Value 3	$D \oplus B$
H			A
I			C
K		Value 4	$D \oplus B$
L			A
M			C
N		Value 5	$D \oplus B$
P			A
R			C
S			
T.			
U			
V			

Running machine values:

6 pin mil connector (male):

connector
(male)

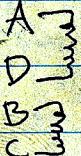
A - A

B - B

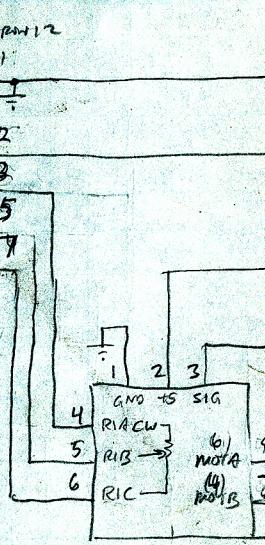
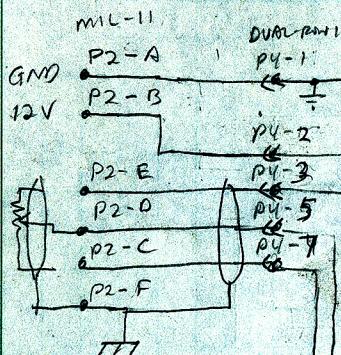
C - C

D - D

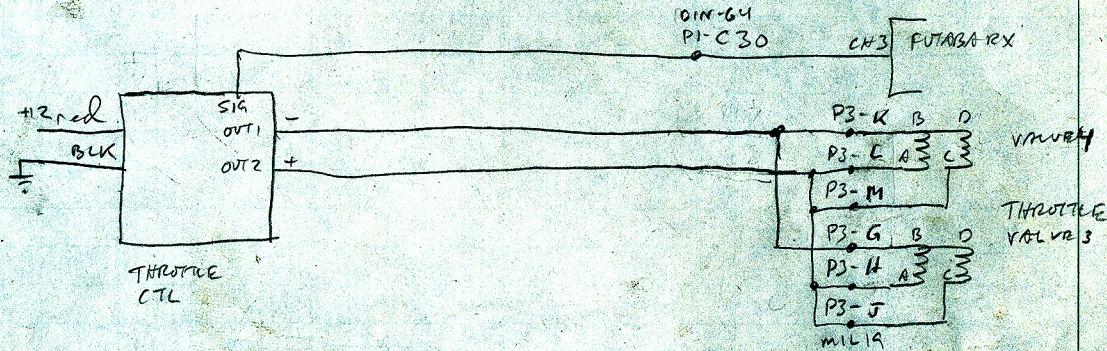
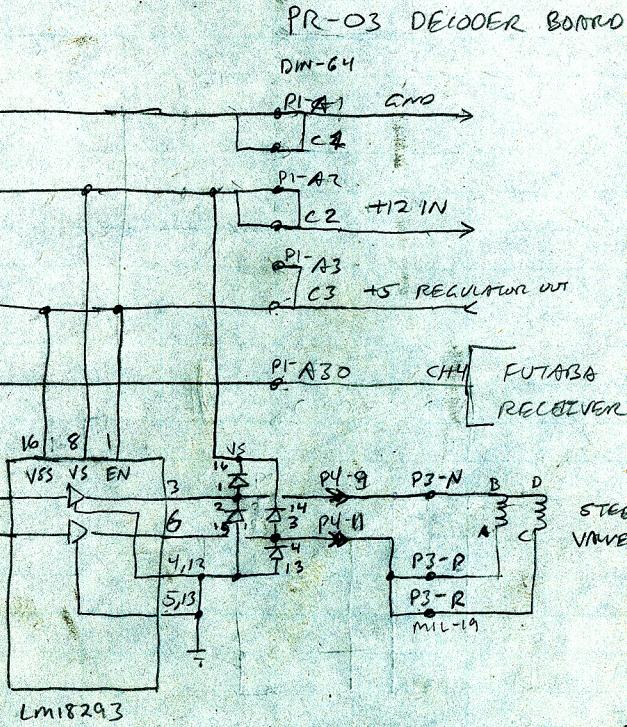
} value labels



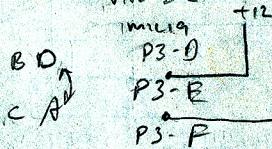
P1 - 64-PIN EUROCARDO MALE
P2 - 11 PIN MIL POWER + FEEDBACK
P3 - 19 PIN MIL VALVE CTL
P4 - 12 PIN DUMARON



FUTABA SERV



UP
A + D
B - Ct
DN
A D+
B + C





**National
Semiconductor**

LM18293 Four Channel Push-Pull Driver

General Description

The LM18293 is designed to drive DC loads up to one amp. Typical applications include driving such inductive loads as solenoids, relays and stepper motors along with driving switching power transistors and use as a buffer for low level logic signals. The four inputs accept standard TTL and DTL levels for ease of interfacing. Two enable pins are provided that also accept the standard TTL and DTL levels. Each enable controls 2 channels and when an enable pin is disabled (tied low), the corresponding outputs are forced to the TRI-STATE® condition. If the enable pins are not connected (i.e., floating), the circuit will function as if it has been enabled. Separate pins are provided for the main power supply (pin 8), and the logic supply (pin 16). This allows a lower voltage to be used to bias up the logic resulting in reduced power dissipation. The chip is packaged in a specially de-

signed 16 pin power DIP. The 4 center pins of this package are tied together and form the die paddle inside the package. This provides much better heat sinking capability than most other DIP packages available. The device is capable of operating at voltages up to 36 volts.

Features

- 1A output current capability per channel
- Pin for pin replacement for L293B
- Special 16 pin power DIP package
- 36 volt operation
- Internal thermal overload protection
- Logical "0" input voltage up to 1.5 volts results in high noise immunity

Typical Connection

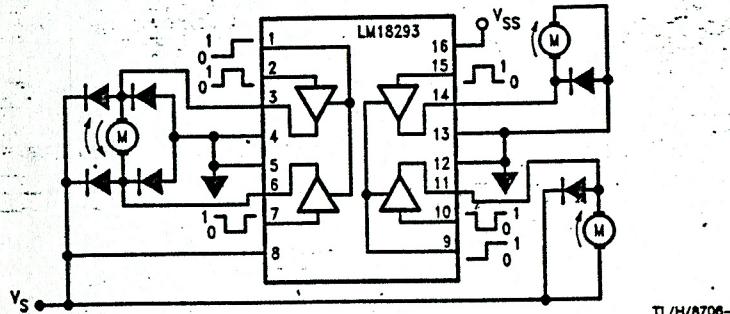


FIGURE 1. Application circuit showing bidirectional and on/off control of a single DC motor using two outputs and unidirectional on/off function of two DC motors using a single output each.

Order Number LM18293N
NS Package Number N16A

Maximum Ratings
Aerospace specified devices are required to meet the National Semiconductor Specification. Distributors for availability and specification information.

Supply Voltage (Vs)

Voltage (Vss)

(V)

Voltage (Ve)

(V)

Voltage (Vd)

(V)

Electrical Characteristics

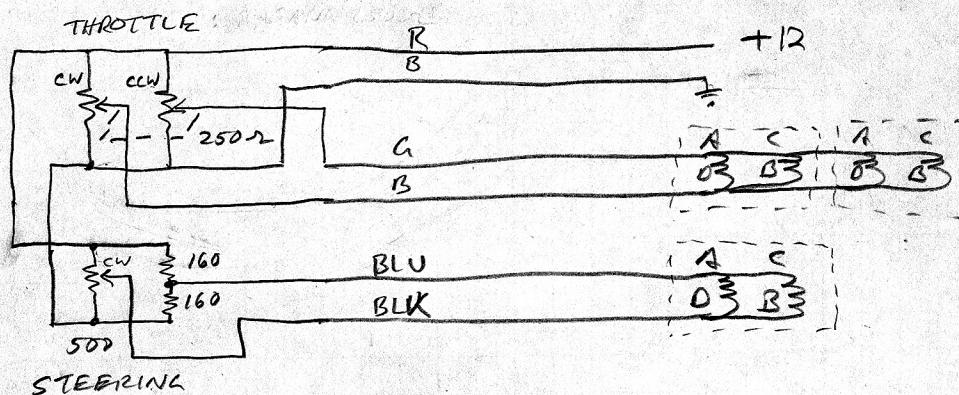
Parameter	
Main Supply (Pin 8)	Maximum
Logic Supply (Pin 16)	Minimum
Total Quiescent Supply Current	$V_1 = L$ $V_1 = H$
Total Quiescent Logic Supply Current (pin 16)	$V_1 = L$ $V_1 = H$
Input Voltage	Min V _i Max V _i Min V _i Max V _i Max V _i
Input Current	$V_1 = L$ $V_1 = H$
Enable Voltage (Pins 1, 9)	Min V _e Max V _e Min V _e Max V _e Max V _e
Enable Current	$V_e = L$ $V_e = H$
Source Saturation Voltage	I_C
Sink Saturation Voltage	I_C
Rise Time	
Fall Time	
Turn-On Delay	
Turn-Off Delay	

Note 1: Tested limits are guaranteed and 100% production tested.
Note 2: Design limits are guaranteed (but not 100% production tested).

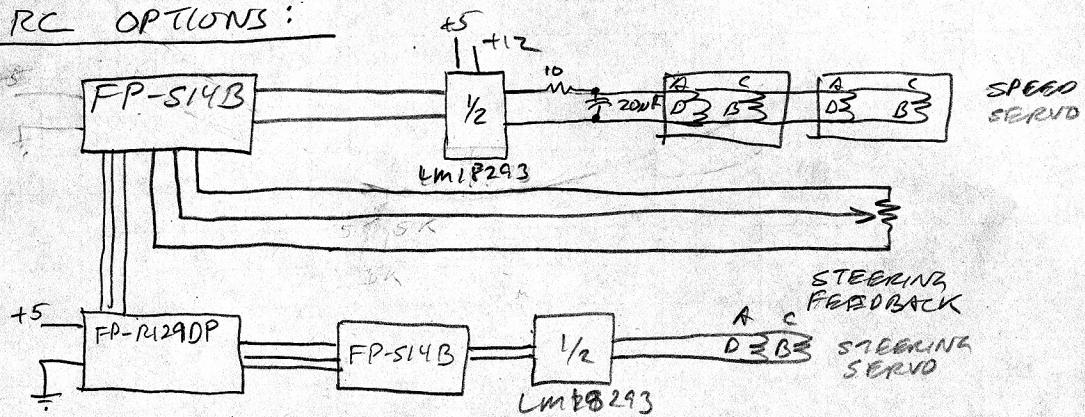
RUNNING MACHINE

RC OPTIONS

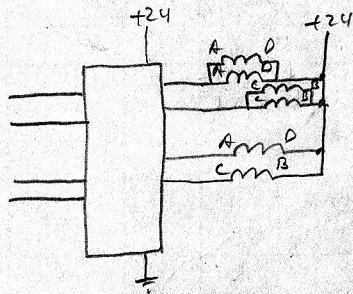
MANUAL CONTROLLER:



RC OPTIONS:

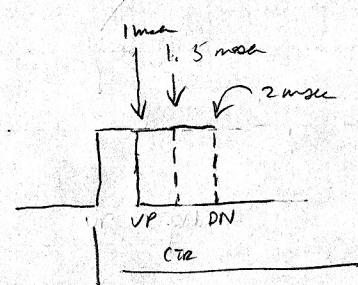


with '293 QUAD HALF-BRIDGE DRIVER CHIP



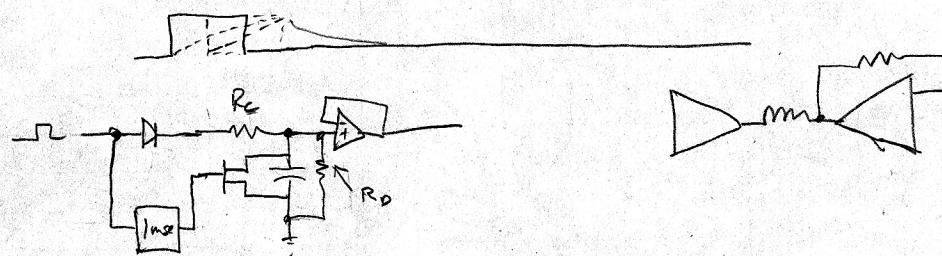
WITH ULN2003

14 msec



BAL 6686
BA 6688 L

1.5 msec

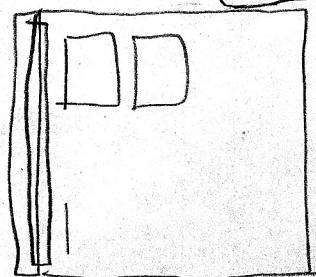
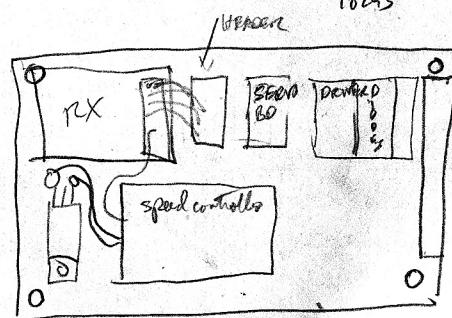
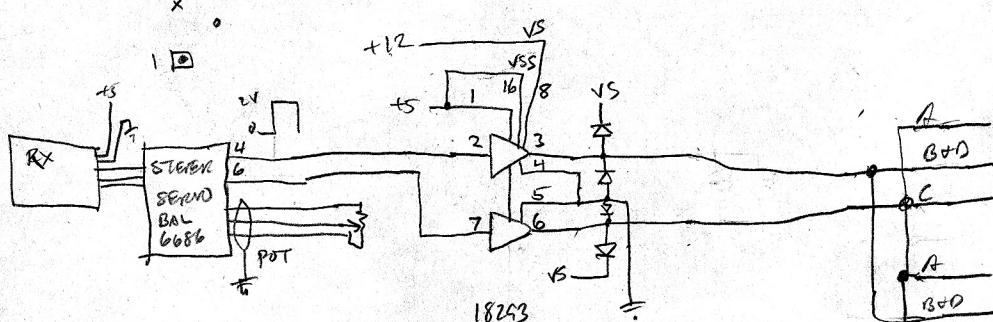


BAL 6686

V+

- x o - STICK DOWN PWM
- o o - STICK UP PWM
- x o

$$,386 - .224 = .162$$



4.9
2.054
113

Screw Machine Update
April 2005
Ralf Burgett

The screw machine now has a Making Things board ("Stepper Block" module) as its controller.

Inputs:

I0 - Futaba Channel 1
I1 - Futaba Channel 2

Outputs :

Q1 - clutch 2
Q2 - clutch 1
Q3 - clutch 4
Q4 - clutch 3

The board decodes the Futaba position and implements the following:

Turn on Q:	
Stick Up	1,4
Stick Down	2,3
Stick Right	1,2
Stick Left	3,4
Stick Up+Right	1
Stick Up+Left	4
Stick Down+Right	2
Stick Down+Left	3

So that with the output wiring given, the clutches are driven as below:

Clutch:	
Stick Up	2,3
Stick Down	1,4
Stick Right	1,2
Stick Left	3,4
Stick Up+Right	2
Stick Up+Left	3
Stick Down+Right	1
Stick Down+Left	4

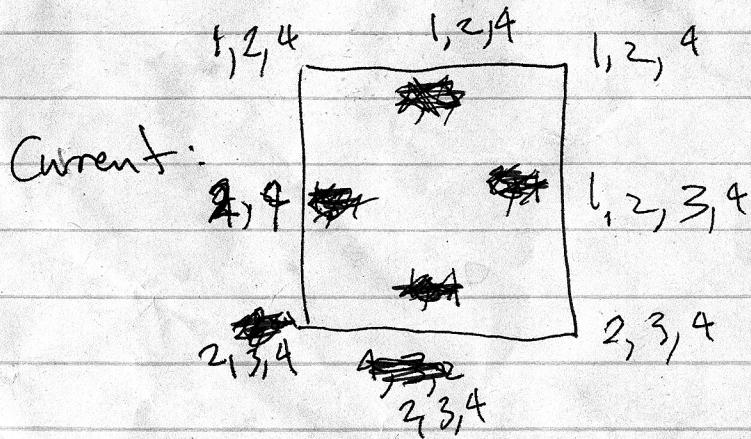
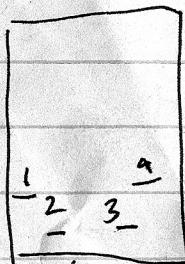
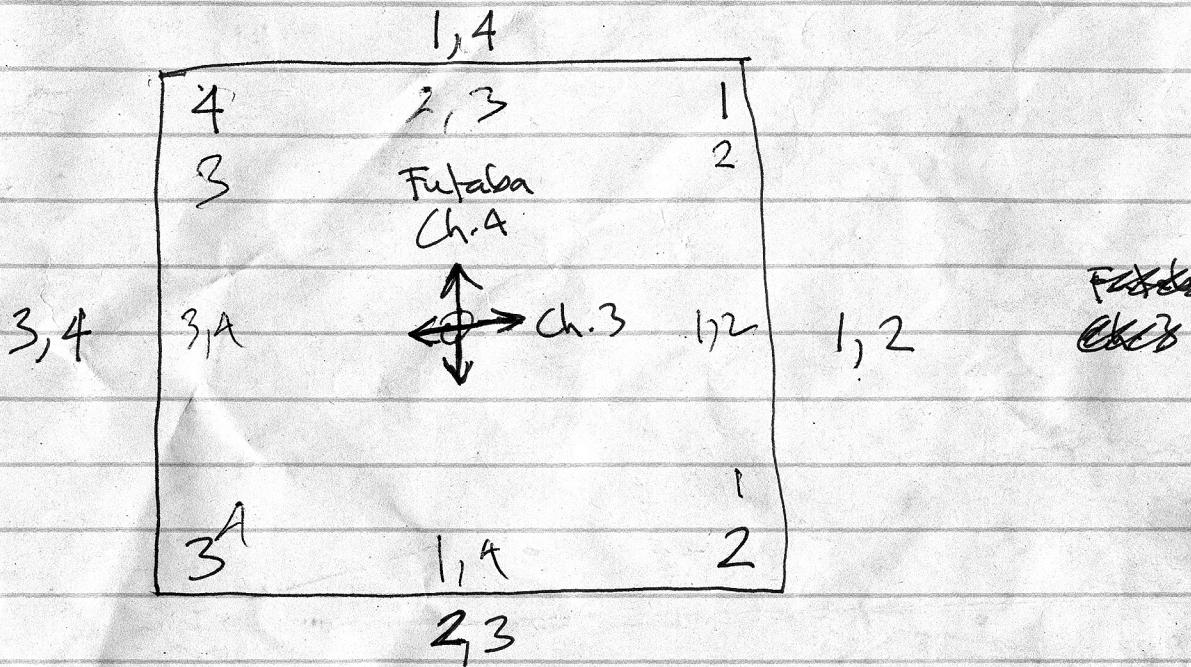
Note that this is different than the mapping in previous versions.

Clutch 3 is currently completely worn and needs to be replaced.

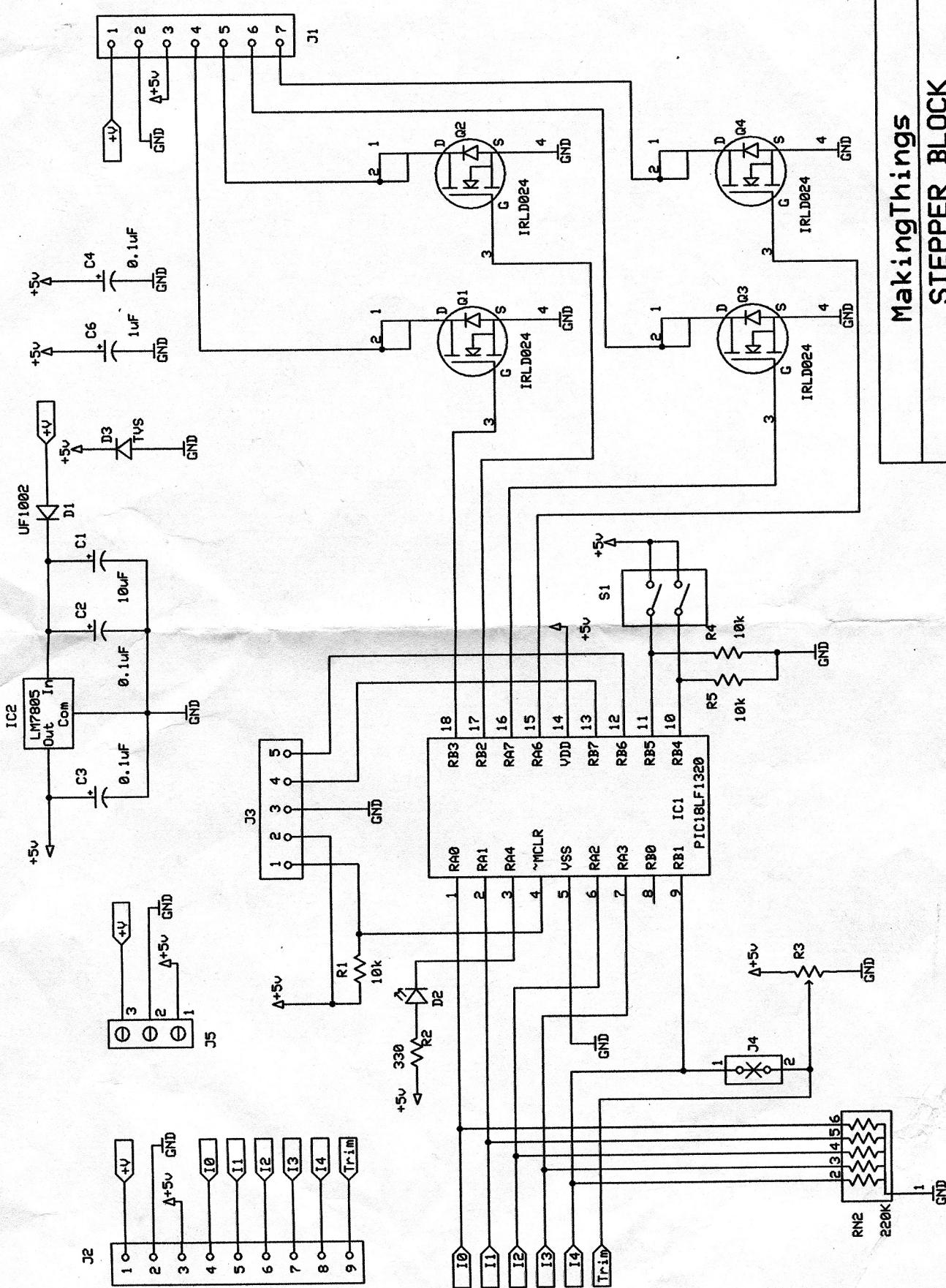
Engine power switch is intermittent and needs to be replaced.

Screw Machine Control Mapping

Clutch numbers:

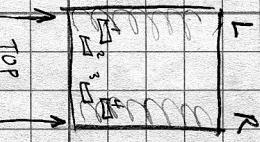


- Ch. 1, 2 to drive 2 FETs each
- Ch. 5 to drive 1 FET
- 70% full joystick travel to turn on, no less.



MakingThings STEPPER BLOCK

David	Rev 1.0
	7/24/2004



<u>CLUTCH</u>	<u>SCREEN</u>	<u>DIRECTION</u>	<u>CLUTCHES</u>
1	L	CW, FWD	RUSH LEFT
2	R	CW, BACKW	RUSH RIGHT
3	L	CW, BACKW	FORWARD
4	R	CW, FWD	BACKWARDS
			TURN F CW
			TURN F CCW
			TURN B CW
			TURN B CCW
			HORN

ILLUSTRATIVE COMBINATIONS

1 & 3
2 & 4

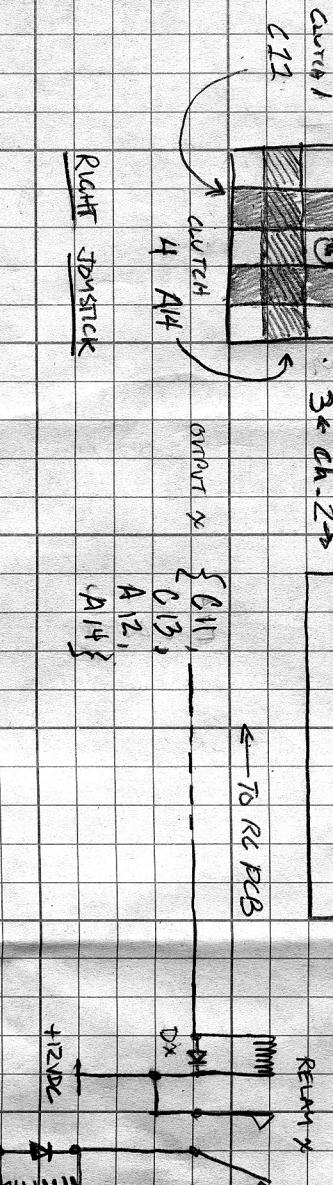
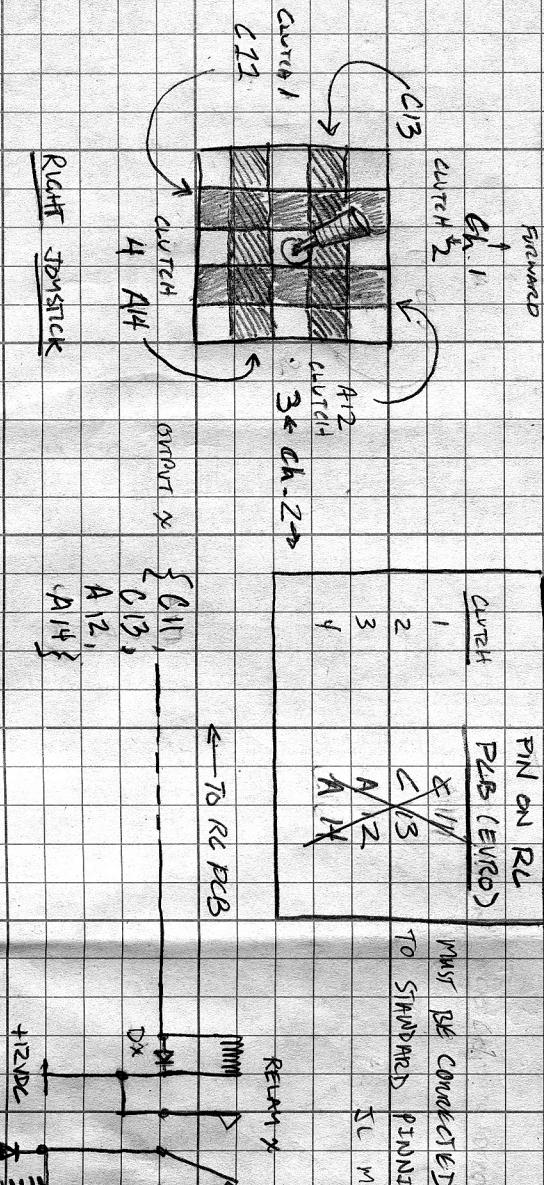
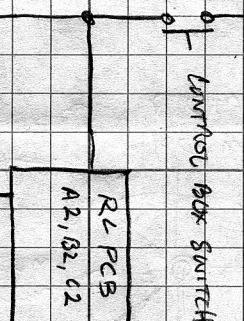
TURN B CCW

HORN

PIN ON RC PCB (EURO)	
1	X // C
2	C // 3
3	A // 2
4	A // 4

TO STANDARD PINNING

JL MAY 21/94



NOTE: THIS DESIGN USES THE RAUL BURGERT RC

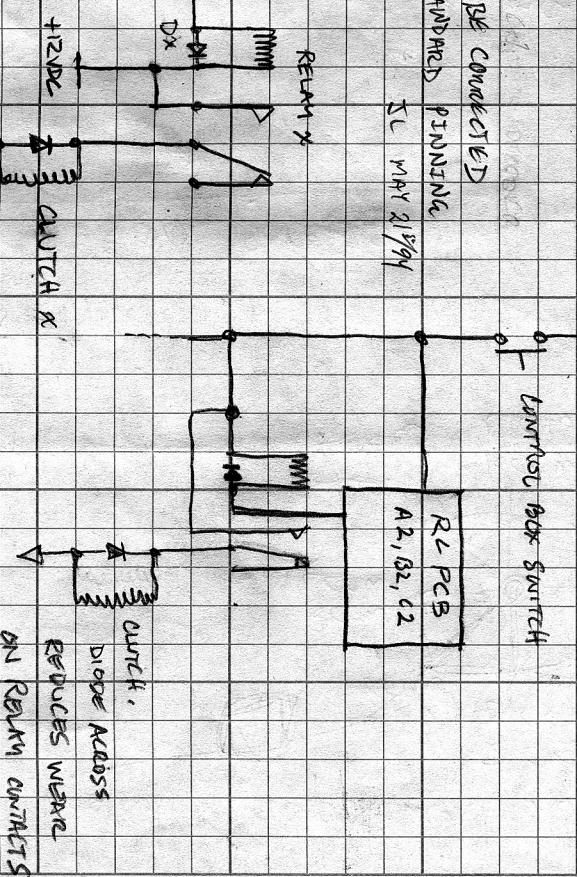
PRINTED CIRCUIT BOARDS. THESE BOARDS HAVE ULN2003 SINKING DIODES TO

DRIVE RELAYS THAT MUST HAVE CLAMPING DIODES.

TOGETHER

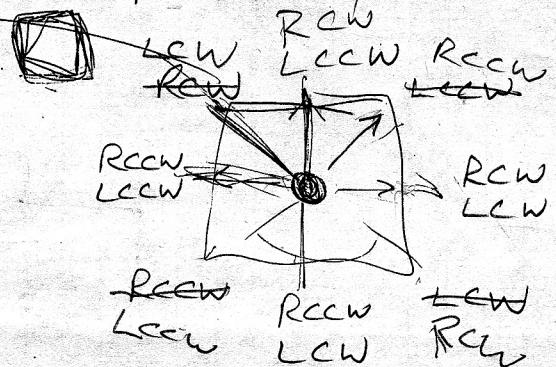
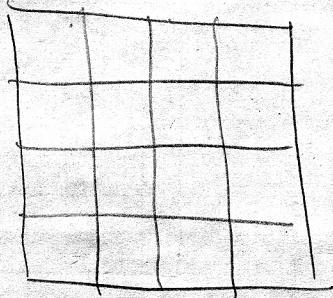
CLUTCH Contact

Schematic



SCREEN MACHINE CONTROL CIRCUIT
M. Fournery 5/1/94

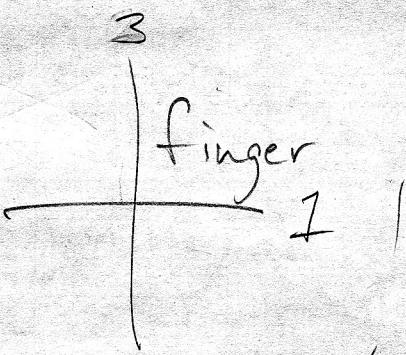
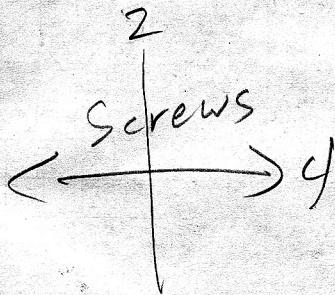
Radio board supplementary ckt



isolate coils for screws

5 comp.
man.
6 attt

1 attachments
7 ↗



2 ⇔ 3 in mode II

**Screw Machine Update
February 1996**

The screw machine's radio board was replaced in Phoenix. The joystick quadrants on the current board are not diode-OR'd together. Doing this made the machine prone to running away if the trim control on the transmitter was accidentally moved.

Screw Logic

$$RCW = U\bar{L}\bar{R} + R\bar{U}$$

$$LCW = UL + R\bar{U}\bar{D} + D\bar{L}\bar{R}$$

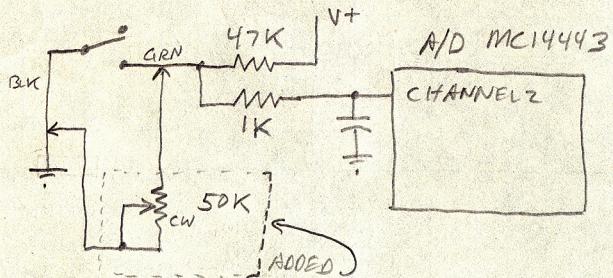
$$RCCW = UR + L\bar{U}\bar{D} + D\bar{L}\bar{R}$$

$$LCCW = U\bar{L}\bar{R} + L\bar{U}$$

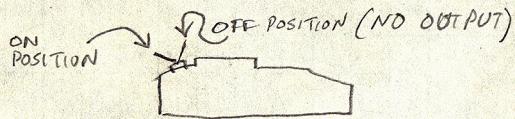
March 1977 V-1 Update:

- Ch. 5 relay (top left of box) connected and working. For props, water injection, etc.
- the second terminal from the left, of the terminal block on the side of the moto-truck motor (under cover) sets the motor speed. Pulling this terminal to ground places a resistance in series with the motor power. The purple wire in the control box can be used for this, but is currently disconnected. Ideally it would be pulled low momentarily when the motor is run, to avoid stalling the motor. Stalling the motor (without the resistor in series) for any length of time will melt the power cable(s).

- Trimpot added: reduces & adjusts "GEAR" switch open voltage.
This allows use of channel 5 on Raybal decoder board.
- NOTE: THIS SIMPLE MOD ALLOWS SELECTION OF ONE (STICK DOWN) OUTPUT ON CHAN.5.

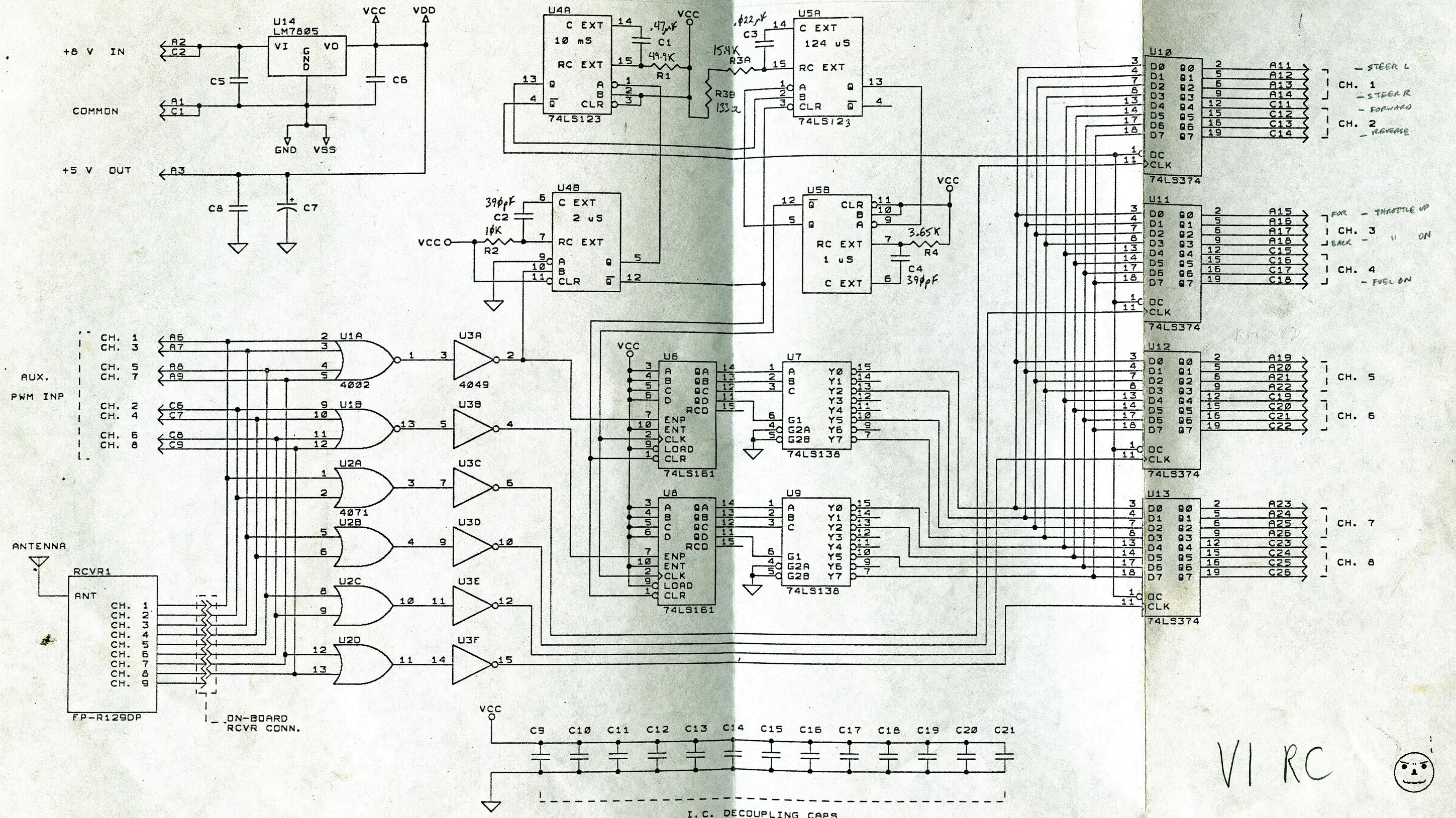


THE TRIMPOT IS TOP PANEL ACCESSIBLE, NEXT TO CH.5 GEAR SWITCH.



March '97 trimpot update:

The new transmitter has a 25K pot. It's located in the D/R switch mounting hole.

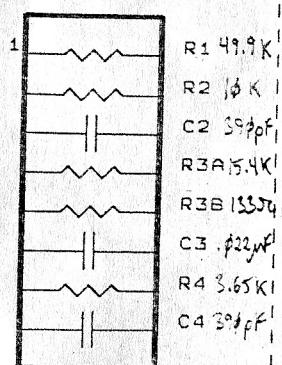
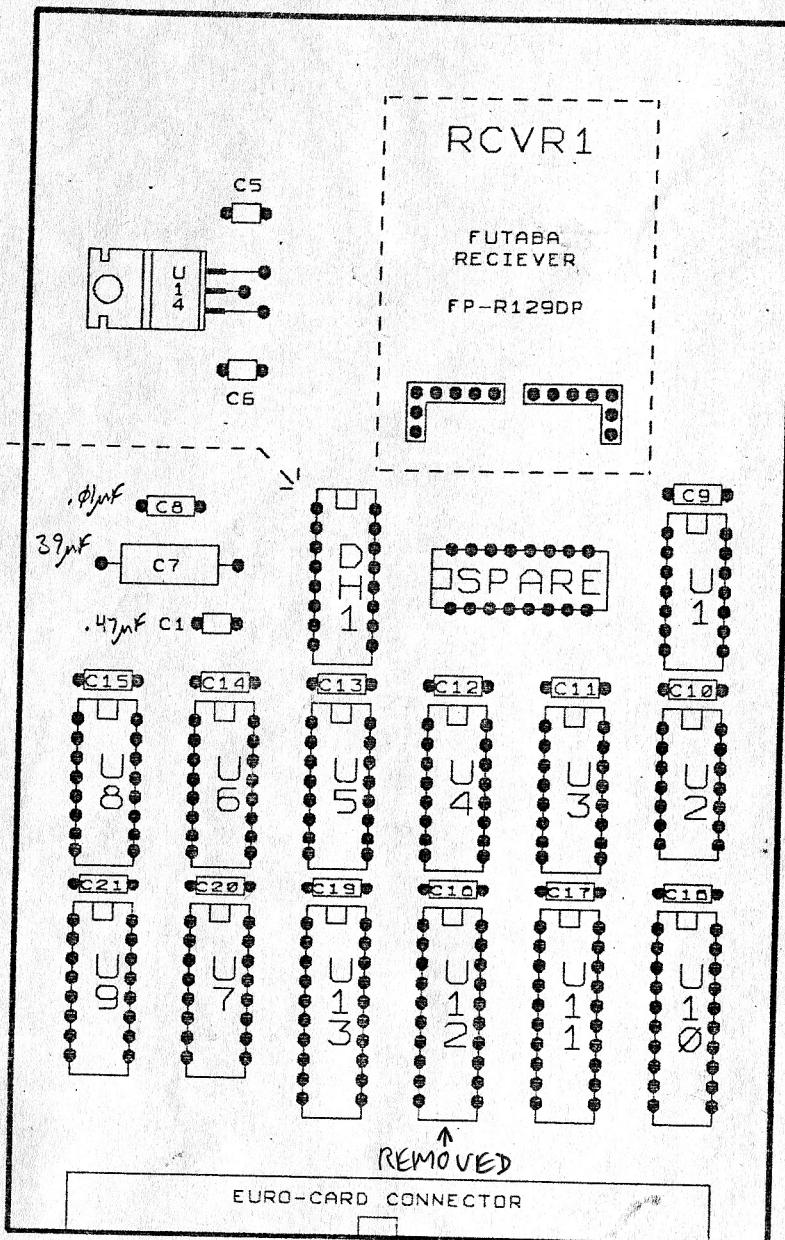


I.C. DECOUPLING CAPS

VI RC

WINGNUT METATRONICS	
Paul Roybal	P.O. BOX 7128
MENLO PARK, CA. 94026	
Title	6 CHANNEL FUTABA TO DIGITAL DECODER
Size	Document Number
6	SRL-PR-001-01
Date	May 29, 1990 Sheet
	1 of 2
REV	5

GENERAL PARTS LAYOUT



VI RC

DETAIL
DH1

WINGNUT METATRONICS		
Paul Roybal P.O. BOX 7128 MENLO PARK, CA. 94025		
Title		
8 CHANNEL FUTABA DECODER		
Size	Document Number	REV
A	SRL-PR-001-01 LAY	0
Date: October 27, 1990 Sheet 2 of 2		

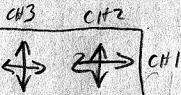
Line	Quantity	Reference	Part			
1	1	C1	0.47	uF	POLY	
2	2	C2,C4	390	pF	MICA	
3	1	C3	22	nF	POLY	
4	1	C5	0.33	uF	CER	
5	2	C6,C8	0.01	uF	CER	
6	1	C7	39	uF 10V	ELECT	
7	13	C9,C10,C11,C12,C13,C14, C15,C16,C17,C18,C19,C20, C21	0.1	uF	CER	
8	1	R1	49.9	K 1%	1/8 W	
9	1	R2	10.0	K 1%	1/8 W	
10	1	R3A	15.4	K 1%	1/8 W	
11	1	R3B	133	OHM 1%	1/8 W	
12	1	R4	3.65	K 1%	1/8 W	
13	1	RCVR1	FP-R129DP (FUTABA RECIEVER)			
14	1	U1	4002			
15	1	U2	4071			
16	1	U3	4049			
17	1	U4,U5	74LS123			
19	2	U6,U8	74LS161			
20	2	U7,U9	74LS138			
21	4	U10,U11,U12,U13	74LS374			
22	1	U14	LM7805			

VI RC RECEIVER

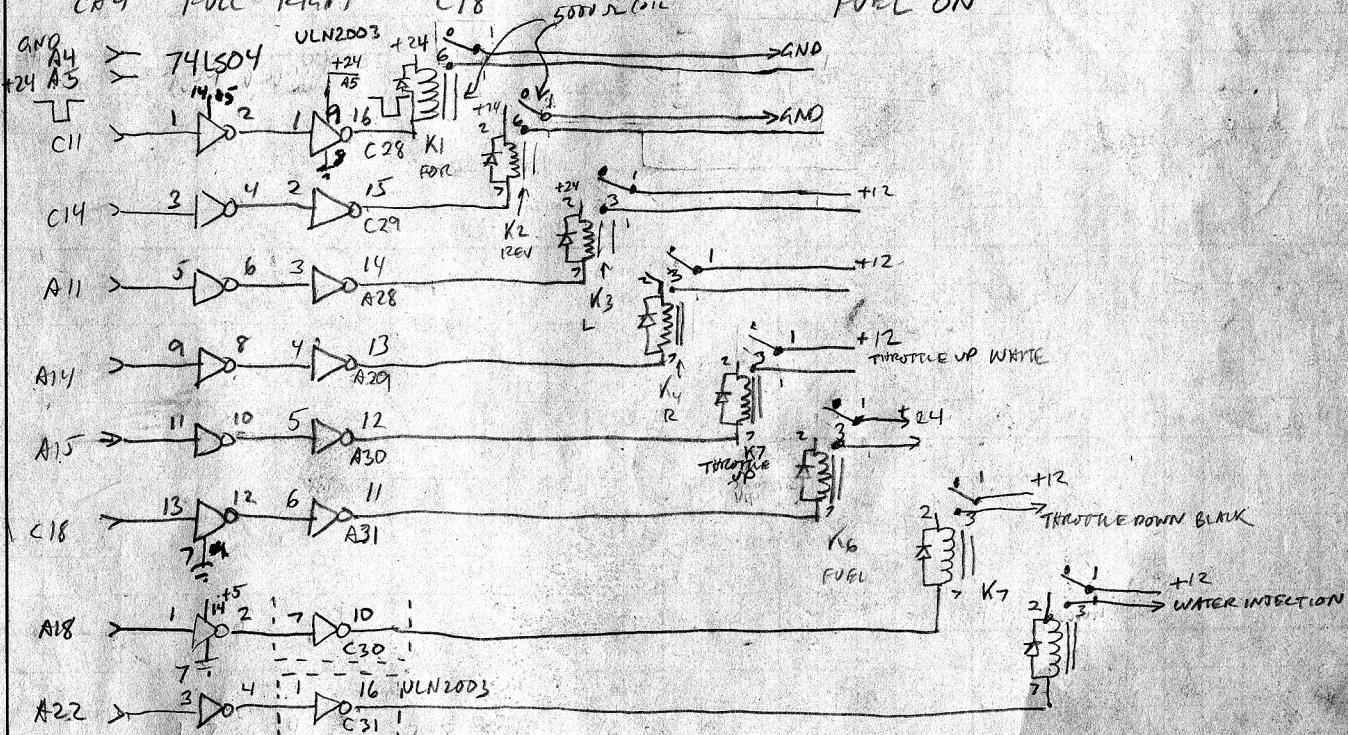
DRIVE REQ:

	CLOSURE TO:	I (Amps)	V Contacts (current)	NOTES
FORWARD	GND	1.7	+2V 3	
Reverse	GND	1.7	12V 5	
Left	12V	3	12V 5	
Right	12V	3	12V 5	STARTER RELAY
Throttle up	12V	3	12V 5	" "
Fuel Clutch	12V	1	12V 5	" "
Throttle down	12V	3	12V	No

22-141 50 SHEETS
22-142 100 SHEETS
22-144 200 SHEETS
MPAD



CHANNEL	POSITION	TTL PIN	D.C. OUT N PIN	FUNCTION
CH 2'	FULL FORWARD	C11	C28	DRIVE { FORWARD
CH 2	FULL REVERSE	C14	C29	REVERSE
CH 5	ON	A22		
CH 1	FULL LEFT	A11	A28	STEER { LEFT
CH 1	FULL RIGHT	A14	A29	RIGHT
CH 3	FULL BACK	A18	C30	THROTTLE DOWN
CH 3	FULL FORWARD	A15	A30	THROTTLE UP
CH 4	FULL RIGHT	C18	A31	FUEL ON



TRANSMITTER NOTES -

- FAILSAFE THROTTLE ("F/S") ON

K1-K7: 24V 80mA DPDT RELAY, 5A contacts

T.S. 1 2 3 4 5 6 | 7 8 9 10 11 12
GND 12 24 VDD 12 110 K1

TRANSMITTER NOTES:

- FAILSAFE THROTTLE ("F/S") ON
- CH 5 SWITCH MOD FOR WATER

